# 3625 AND 3626

# PowerHall<sup>®</sup> SENSOR IC/DRIVERS FOR BRUSHLESS DC MOTORS

PWM/Io ADJUST 1 8 TACH OUT X INTERNAL CONNECTION 2 IC GROUND 7 LOGIC CONTROL CONNECTION 3 IC 0UT₄ 6 OUTB SUPPLY Dwg. No. PP-012 **ABSOLUTE MAXIMUM RATINGS** Supply Voltage, V<sub>CC</sub> . . . . . . . . See Table Output Current, I<sub>0</sub>..... See Table PWM Current Control Voltage, V<sub>ADJ</sub> ..... See Applications Magnetic Flux Density, B Unlimited Package Power Dissipation, See Graph Ρ<sub>D</sub>..... Operating Temperature Range, T<sub>A</sub>......**-20°C to +85°C** Storage Temperature Range, UDN3625M **UDN3626M** 14 V 26 V V<sub>CC(MAX)</sub> 1.0 A 0.45 A I<sub>O(CONT)</sub> 1.6 A 0.75 A I<sub>O(PEAK)</sub> Output current rating will be limited by ambient temperature, supply voltage, and

ambient temperature, supply voltage, and duty cycle. Under any set of conditions, do not exceed a junction temperature of +150°C. Requiring a minimum of external components, the UDN3625M and UDN3626M are monolithic ICs that provide single-chip control and direct drive solutions for many small, single-phase, unipolar brushless dc motor applications. Integrated into the ICs are a highsensitivity Hall-effect element, control and commutating logic, a stable voltage regulator, extensive self and system protective functions, and two high-current saturated npn outputs. Both ICs include thermal shutdown, output over-current limiting, and output transient protection/ flyback diodes. The UDN3625M is nominally for 12 V motor applications while the UDN3626M is better suited to 24 V motors.

Output over-current limiting (relating to startup surge or a locked rotor condition) and short-circuit protection are provided by an internal current-sense resistor. The maximum (default) output load current is typically 1.3 A for the UDN3625M or 600 mA for the UDN3626M, but may be decreased by user selection of an external low-wattage resistor.

A separate low-level output provides tachometer capability for motor speed control or sensing a locked rotor condition. With appropriate external logic, pulse-width modulated (PWM) speed control can be accomplished at the output current adjust pin.

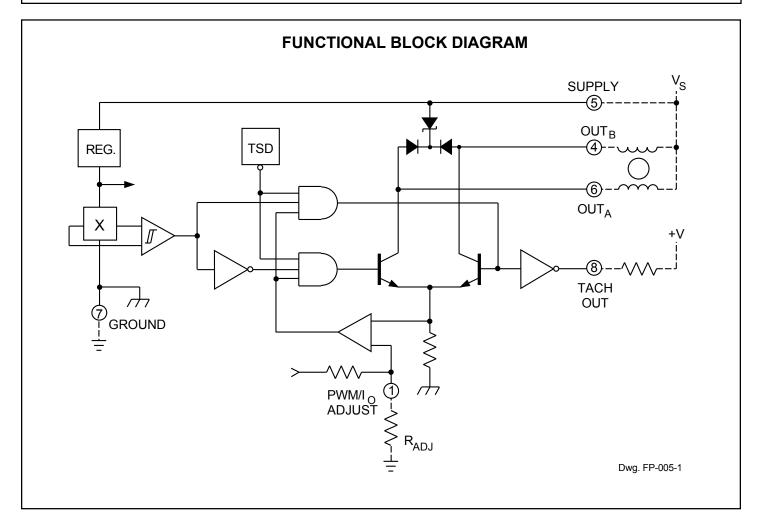
These sensor ICs/drivers are supplied in an 8-pin mini-DIP plastic package with a copper leadframe for increased package power handling capability.

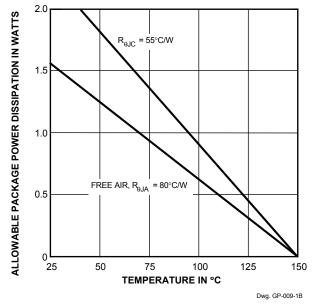
## FEATURES

- 900 mA/12 V or 400 mA/24 V Operation
- Speed-Control (PWM) Capability
- Locked-Rotor Indication
- Minimum External Components
- Over-Current Protected
- Thermal Protection
- Enhanced Reliability
- Reduced Cost

Always order by complete part number, e.g., **UDN3625M**. See Maximum Ratings at left.







#### **TRUTH TABLE**

Mag. Field	PWM/I <sub>O</sub> ADJ	OUTA	OUTB
> +B <sub>OP</sub>	Open	Low	High
> -B <sub>OP</sub>	Open	High	Low
Any	<0.3 V	High	High



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# ELECTRICAL CHARACTERISTICS at $T_A = +25^{\circ}C$ , $R_{ADJ} = \infty$ , Over Operating Voltage Range (unless otherwise noted).

Characteristic	Symbol	Test Conditions	Limits			
			Min.	Тур.	Max.	Units
UDN3625M	-					
Operating Voltage Range	Vs		6.5	12	14	V
Output Leakage Current	Ι <sub>Ο</sub>	$V_{CC} = V_O = 24 V$		_	100	μA
Output Breakdown Voltage	V <sub>(BR)CEX</sub>	V <sub>CC</sub> = 24 V, I <sub>O</sub> = 10 mA*	34	_		V
Output Saturation Voltage	V <sub>O(SAT)</sub>	I <sub>O</sub> = 450 mA, V <sub>S</sub> = 6.5 V		0.25	0.4	V
		I <sub>O</sub> = 900 mA, V <sub>S</sub> = 12 V	_	0.5	0.8	V
Output Current Limit	I <sub>O</sub> MAX	$R_{ADJ} = \infty, V_O = 2 V$	1.0	1.3	1.6	A
Output Clamp Voltage (Test Fig. 1)	V <sub>CL</sub>	$I_{CL} = 10 \text{ mA}, V_{CC} = 0$	11	12	13	V
		I <sub>CL</sub> = 450 mA=, V <sub>CC</sub> = 0	13	14	15	V
Output Switching Time	t <sub>PLH</sub>	50% $V_{ADJ}$ to $V_{O}$ = 3.0 V, $I_{O}$ = 450 mA	_	_	5.0	μs
Supply Current (Test Fig. 2)	I <sub>CC</sub>	$R_{ADJ} = \infty$ , $V_S = 14$ V, One Output ON	_	30	38	mA
		$R_{ADJ}$ = 0 $\Omega$ , $V_{S}$ = 14 V, Outputs OFF	_	8.0	10	mA
UDN3626M	•	•				
Operating Voltage Range	Vs		‡	24	26	V
Output Leakage Current	Ι <sub>Ο</sub>	$V_{CC} = V_O = 30 V$		—	100	μA
Output Breakdown Voltage	V <sub>(BR)CEX</sub>	V <sub>CC</sub> = 30 V, I <sub>O</sub> = 10 mA*	50	_	_	V
Output Saturation Voltage	V <sub>O(SAT)</sub>	I <sub>O</sub> = 250 mA, V <sub>S</sub> = 6.5 V	-	0.15	0.3	V
		I <sub>O</sub> = 400 mA, V <sub>S</sub> = 24 V	_	0.3	0.5	V
Output Current Limit	I <sub>O</sub> MAX	$R_{ADJ} = \infty, V_O = 2 V$	450	600	750	mA
Output Clamp Voltage (Test Fig. 1)	V <sub>CL</sub>	$I_{CL} = 10 \text{ mA}, V_{CC} = 0$	17	18	20	V
		I <sub>CL</sub> = 250 mA, V <sub>CC</sub> = 0	19	20	22	V
Output Switching Time	t <sub>PLH</sub>	50% $V_{ADJ}$ to $V_O$ = 3.0 V, $I_O$ = 250 mA	_		5.0	μs
Supply Current	I <sub>CC</sub>	$R_{ADJ} = \infty$ , $V_S = 26$ V, One Output ON	_	_	24	mA
(Test Fig. 2)		$R_{ADJ} = 0 \Omega$ , $V_S = 26 V$ , Outputs OFF	_	8.0	10	mA

\* l<sub>0</sub> is almost entirely Zener clamp current.

† Pulse test.

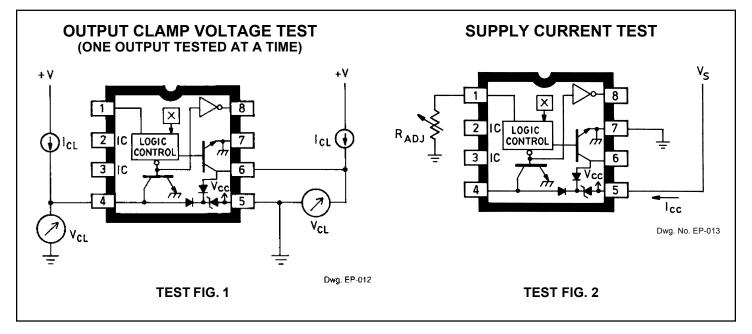
‡ Dependent on value of external series Zener diode (see Applications), 6.5 V without a Zener diode.

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#### ELECTRICAL CHARACTERISTICS continued

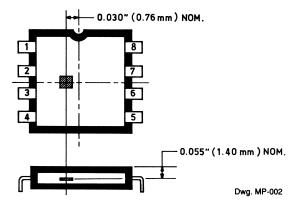
	Symbol	Test Conditions		Limits			
Characteristic			Min.	Тур.	Max.	Units	
Both	<u>.</u>		•			•	
Operate Point§	B <sub>OP</sub>		—	±100	±150	G	
Hysteresis	B <sub>H</sub>		_	200	_	G	
Output Current Limit Ratio	I <sub>O</sub> /I <sub>O</sub> MAX	$R_{ADJ}$ = 39 k $\Omega$	_	0.75	_	_	
		$R_{ADJ}$ = 17 k $\Omega$	_	0.50	_	_	
		R <sub>ADJ</sub> = 10 kΩ	_	0.25	_	_	
PWM Control Current	I <sub>ADJ</sub>	V <sub>ADJ</sub> = 0	_	-350	-500	μA	
Tach Output Leak. Current	Ι <sub>Τ</sub>	V <sub>T</sub> = 14 V	—	_	10	μA	
Tach Output Sat. Voltage	V <sub>T(SAT)</sub>	I <sub>T</sub> = 750 μA	_	0.2	0.4	V	
Thermal Shutdown	TJ			165	_	°C	
Thermal Hysteresis	$\Delta T_{J}$			10	_	°C	

§ Magnetic flux density is measured at most sensitive area of device, nominally located 0.055" (1.40 mm) below the top of the package.





#### ELEMENT LOCATION



#### APPLICATIONS INFORMATION

**Power Dissipation.** Care should be taken in evaluating the package power dissipation of these devices. Total power dissipated by the device will consist of power due to the internal regulator, logic and drive circuitry ( $I_{CC} \times V_{CC}$ ), power due to the output drivers ( $I_{O} \times V_{O(SAT)}$ ), and power due to the clamp circuitry ( $I_{CL} \times V_{CL} \times duty$  cycle).

For example:

 $I_{CC} \ge V_{CC} = 38 \text{ mA} \text{ (max)} \ge 14 \text{ V} = 532 \text{ mW} \text{ (max)}$ 

 $I_O \ge V_{O(SAT)} = 450 \text{ mA } \ge 0.4 \text{ V} \text{ (max)} = 180 \text{ mW} \text{ (max)}$ 

 $I_{CL} \times V_{CL} \times duty \ cycle = 450 \ mA \times 15 \ V \ (max) \times 1\% = 67.5 \ mW \ (max)$ 

Total package power dissipation = 779.5 mW (max)

Some of the power dissipated by the device  $(I_{CC} \times V_{CC})$  can be reduced by inserting a Zener diode in the supply line ( $Z_A$  in the figure). Note that the voltage at the V<sub>CC</sub> pin under worst-case conditions must be greater than the minimum operating voltage (6.5 V).

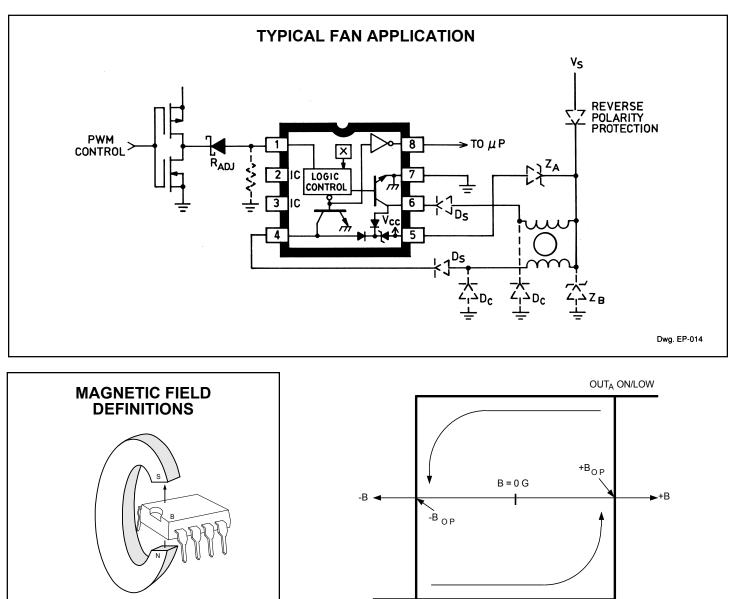
**Transient Protection.** A note of caution concerns negative (below ground) excursions of the outputs. In application, the coupling of the two motor windings can provide for just such a case. Reducing the coupling between windings will help, but ground clamp diodes or diodes in series with the motor windings might be required ( $D_C$  or  $D_S$  in the figure). Most small brushless motors will not require these diodes.

System requirements usually utilize a diode type of reverse-polarity protection. If series diode protection is used with an inductive load (the usual fan application), a Zener clamp between  $V_S$  and ground ( $Z_B$  in the figure) is required. The Zener voltage must be greater than the supply voltage but less than the rated maximum allowable supply voltage.

With diode reverse-polarity protection, a high-impedance supply, or a switched supply line, high-voltage spikes will be generated (especially with high-current or high-inductance loads) during normal operation, coasting, or immediately after turn-off. In these situations, a Zener clamp ( $Z_B$ ) from  $V_S$  to ground will be required.

**Over-Current Adjust Pin.** The over-current limit may be reduced from the default value ( $I_O MAX$ ) by selection of an external resistor ( $R_{ADJ}$ ) at the PWM/ $I_O ADJ$  pin.

The external overcurrent adjust and the thermal shutdown are commoned at the PWM/l<sub>o</sub> ADJ pin and tieing it to  $V_{CC}$  will disable the thermal shutdown. PWM current/speed control can be performed at the PWM/l<sub>o</sub> ADJ pin from a standard totem-pole logic output with a series Schottky diode (1N5818,1N5819, or equivalent) or by pulling it low through an open-collector transistor (no pull-up resistor). PWM/l<sub>o</sub> ADJ input voltages greater than 0.3 V are not recommended and may create an unstable operating condition.



OUT<sub>B</sub> ON/LOW

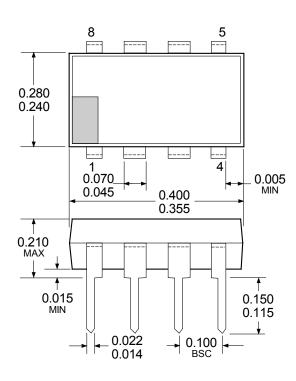
Dwg. GP-008

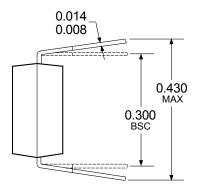
By accepted magnetic convention, lines of flux emanate from the north pole of a magnet and enter the south pole.



Dwg. AP-001-1

# **Dimensions in Inches** (controlling dimensions)





Dwg. MA-001-8A in

NOTES: 1. Leads 1, 4, 5, and 8 may be half leads at vendor's option.

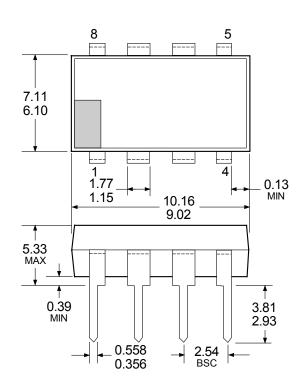
2. Lead thickness is measured at seating plane or below.

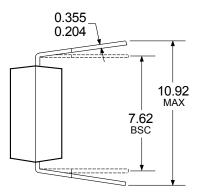
3. Lead spacing tolerance is non-cumulative.

4. Exact body and lead configuration at vendor's option

within limits shown.

#### Dimensions in Millimeters (for reference only)





Dwg. MA-001-8A mm

The products described herein are manufactured under one or more of the following U.S. patents: 5,045,920; 5,264,783; 5,442,283; 5,389,889; 5,581,179; 5,517,112; 5,619,137; 5,621,319; 5,650,719; 5,686,894; 5,694,038; 5,729,130; 5,917,320; and other patents pending.

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