ACS730LLC

Automotive-Grade, 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in Small Footprint SOIC8 Package

Not for New Design

The ACS730LLC is in production but has been determined to be NOT FOR NEW DESIGN. This classification indicates that sale of this device is currently restricted to existing customer applications. The device should not be purchased for new design applications because obsolescence in the near future is probable. Samples are no longer available.

Date of status change: March 14, 2025

Recommended Substitutions:

For existing customer transition, and for new customers or new applications, refer to CT428, ACS37030, ACS37032, or ACS37035.

NOTE: For detailed information on purchasing options, contact your local Allegro field applications engineer or sales representative.

Allegro MicroSystems reserves the right to make, from time to time, revisions to the anticipated product life cycle plan for a product to accommodate changes in production capabilities, alternative product availabilities, or market demand. The information included herein is believed to be accurate and reliable. However, Allegro MicroSystems assumes no responsibility for its use; nor for any infringements of patents or other rights of third parties which may result from its use.

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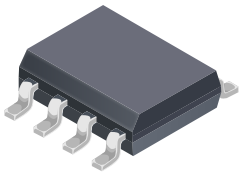
FEATURES AND BENEFITS

- AEC-Q100 automotive qualified
- Industry-leading noise performance with greatly improved bandwidth through proprietary amplifier and filter design techniques
- High bandwidth 1 MHz analog output
- Patented integrated digital temperature compensation circuitry allows high accuracy over temperature in an open loop sensor
- 1.2 mΩ primary conductor resistance for low power loss and high inrush current withstanding capability
- Small footprint, low-profile SOIC8 package suitable for space-constrained applications
- Integrated shield virtually eliminates capacitive coupling from current conductor to die due to high dV/dt voltage transients
- 5 V single supply operation
- Output voltage proportional to AC or DC current
- Factory-trimmed sensitivity and quiescent output voltage for improved accuracy
- High PSRR for noisy environments

PACKAGE: 8-Pin SOIC (suffix LC)



Not to scale

TÜV America
Certificate Number:
U8V 14 11 54214 032
CB 14 11 54214 031

CB Certificate Number:
US-22334-A2-UL

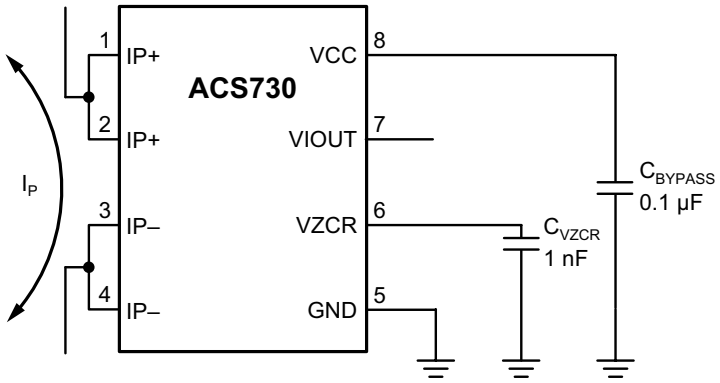
DESCRIPTION

The Allegro™ ACS730 provides economical and precise solutions for AC or DC current sensing in automotive, industrial, commercial, and communications systems. The device package allows for easy implementation by the customer. Typical applications include on-board chargers (OBC), DC-to-DC converters, motor control, load detection and management, switched-mode power supplies, and overcurrent fault protection.

The device consists of a precise, low-offset, linear Hall sensor circuit with a copper conduction path located near the surface of the die. Applied current flowing through this copper conduction path generates a magnetic field which is sensed by the integrated Hall IC and converted into a proportional voltage. Device accuracy is optimized through the close proximity of the magnetic field to the Hall transducer. A precise, proportional voltage is provided by the Hall IC, which is programmed for accuracy after packaging. The output of the device has a positive slope when an increasing current flows through the primary copper conduction path (from pins 1 and 2, to pins 3 and 4), which is the path used for current sensing. The internal resistance of this conductive path is typically 1.2 mΩ, providing low power loss.

The terminals of the conductive path are electrically isolated from the sensor leads (pins 5 through 8). This allows the ACS730 current sensor to be used in high-side current sense applications without the use of high-side differential amplifiers or other costly isolation techniques.

The ACS730 is provided in a small, low-profile surface-mount SOIC8 package. The leadframe is plated with 100% matte tin, which is compatible with standard lead (Pb) free printed circuit board assembly processes. Internally, the flip-chip device is considered Pb-free. However, the solder bump connections are available in a Pb-free or high-temperature Pb-based option. Part numbers followed by "-S" are manufactured with tin-silver-based solder bumps, making these parts Pb-free compliant without the use of RoHS exemptions. Part numbers followed by "-T" are manufactured with Pb-based solder bumps using allowed RoHS exemptions.



Typical Application

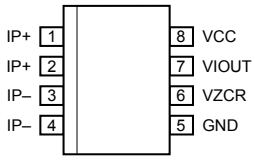
The ACS730 outputs an analog signal, V_{IOUT}, that varies linearly with the bidirectional AC or DC primary sensed current, I_P, within the range specified.

SELECTION GUIDE

Part Number	Optimized Range, I_P (A)	Sensitivity [1], Sens(Typ) (mV/A)	T_A (°C)	Packing
-S VARIANT [2]				
ACS730LLCTR-50AB-S	±50	40		
-T VARIANT [3]				
ACS730LLCTR-40AU-T	40	100	–40 to 150	Tape and reel, 3000 pieces per reel
ACS730LLCTR-50AB-T	±50	40		

[1] Measured at V_{CC} = 5 V.
[2] "-S" denotes lead-free construction with tin-silver-based solder bumps.
[3] "-T" denotes Pb-contained construction with Pb-based solder bumps. Operating performance of "-T" and "-S" devices are identical. "-T" devices are RoHS compliant using allowed exemptions provided in Annex III and IV of Directive 2011/65/EU [Exemptions 7(a), 15, 15(a), as applicable].

Pinout Diagram and Terminal List Table



Package LC, 8-Pin SOICN Pinout Diagram

Terminal List Table

Number	Name	Description
1, 2	IP+	Terminals for current being sensed; fused internally
3, 4	IP–	Terminals for current being sensed; fused internally
5	GND	Signal ground terminal
6	VZCR	Zero current reference
7	VIOUT	Analog output signal
8	VCC	Device power supply terminal

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ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Notes	Rating	Units
Supply Voltage	V_{CC}		6	V
Reverse Supply Voltage	$V_{CC(R)}$		–0.1	V
Output Voltage	V_{IOUT}		6	V
Reverse Output Voltage	$V_{IOUT(R)}$		–0.1	V
Zero Current Reference Voltage	V_{ZCR}		20	V
Reverse Zero Current Reference Voltage	$V_{ZCR(R)}$		–0.1	V
Operating Ambient Temperature	T_A	Range L	–40 to 150	°C
Junction Temperature	$T_{J(max)}$		165	°C
Storage Temperature	T_{stg}		–65 to 170	°C

ISOLATION CHARACTERISTICS

Characteristic	Symbol	Notes	Value	Units
Withstand Strength [1][2]	V_{ISO}	Agency rated for 60 seconds per UL 62368-1 (edition 3)	2400	V_{RMS}
Working Voltage for Basic Isolation [2]	V_{WVBI}	Maximum approved working voltage for basic (single) isolation according to UL 62368-1 (edition 3)	420	V_{PK} or VDC
			297	V_{RMS}
Impulse Withstand [2]	$V_{IMPULSE}$	Tested ±5 pulses at 2/minute in compliance to IEC 61000-4-5, 1.2 µs (rise) / 50 µs (width)	2500	V_{RMS}
Clearance	D_{cl}	Minimum distance through air from IP leads to signal leads	4	mm
Creepage	D_{cr}	Minimum distance along package body from IP leads to signal leads	4	mm
Distance Through Insulation	DTI	Minimum internal distance through insulation	44	µm
Comparative Tracking Index	CTI	Material Group II	400 to 599	V

[1] Production tested for 1 second in accordance with UL 62368-1 (edition 3).
[2] Certification pending.

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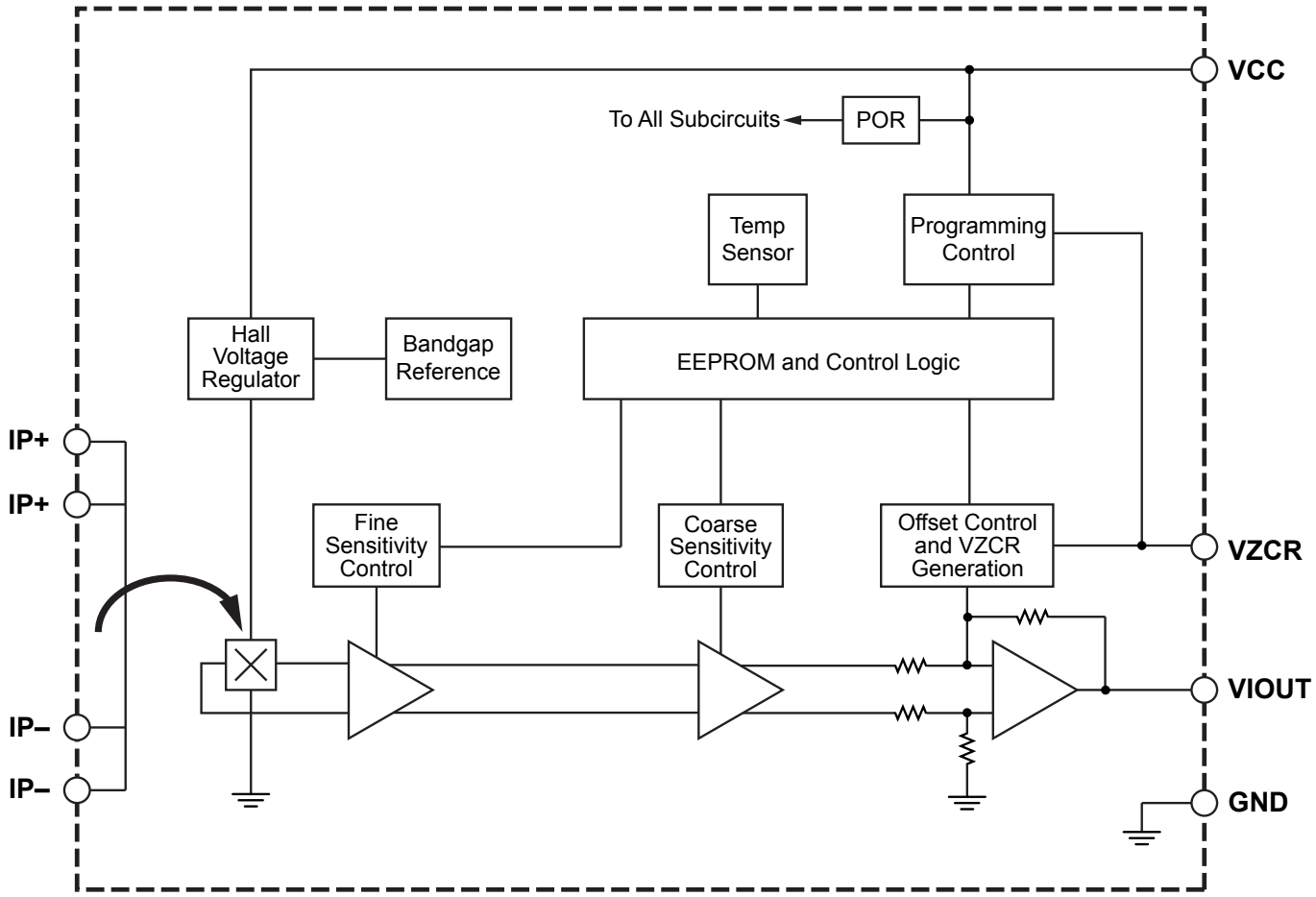
Functional Block Diagram



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COMMON ELECTRICAL CHARACTERISTICS [1]: Valid over full range of T_A, V_{CC} = 5 V, unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Supply Voltage	V_{CC}		4.5	5	5.5	V
Supply Current	I_{CC}	V_{CC} = 5 V, output open	–	17	25	mA
Power-On Time	t_{PO}	T_A = 25°C	–	150	–	µs
Output Capacitance Load	C_L	VIOUT to GND	–	–	0.47	nF
Reference Capacitance Load	C_{VZCR}	VZCR to GND	–	–	1	nF
Output Resistive Load	R_L	VIOUT to GND, VIOUT to VCC	10	–	–	kΩ
Reference Resistive Load	R_{VZCR}	VIOUT to GND, VZCR to VCC	10	–	–	kΩ
Output High Saturation Voltage [2]	V_{OH}	VIOUT, VZCR	V_{CC} – 0.4	V_{CC} – 0.3	–	V
Output Low Saturation Voltage [2]	V_{OL}	VIOUT, VZCR	–	0.1	0.2	V
Primary Conductor Resistance	R_{IP}	T_A = 25°C	–	1.2	–	mΩ
Magnetic Coupling Factor	MCF	T_A = 25°C	–	10	–	G/A
Rise Time	t_r	T_A = 25°C, C_L = 0.47 nF, 1 V step on output	–	0.6	–	µs
Propagation Delay	t_{pd}	T_A = 25°C, C_L = 0.47 nF, 1 V step on output	–	0.2	–	µs
Response Time	$t_{RESPONSE}$	T_A = 25°C, C_L = 0.47 nF, 1 V step on output	–	0.7	–	µs
Internal Bandwidth	BW	Small signal –3 dB; C_L = 0.47 nF	–	1	–	MHz
Noise Density	I_{ND}	Input-referenced noise density; T_A = 25°C, C_L = 0.47 nF	–	40	–	µA/√(Hz)
Noise	I_N	Input-referenced noise density; T_A = 25°C, C_L = 0.47 nF	–	50	–	mA_{RMS}
Power Supply Rejection Ratio	PSRR	0 to 200 Hz, 100 mV pk-pk ripple on V_{CC}, I_P = 0 A, VIOUT and VZCR	–	35	–	dB
Sensitivity Power Supply Rejection Ratio	SPSRR	DC, $V_{CC}(min) < V_{CC} < V_{CC}(max)$, $I_P = I_{PR}(max)$	–	15	–	dB
Offset Power Supply Rejection Ratio	OPSRR	DC, $V_{CC}(min) < V_{CC} < V_{CC}(max)$	–	30	–	dB
Output Source Current	$I_{OUT(src)}$	VIOUT shorted to GND	–	5.5	–	mA
Output Sink Current	$I_{OUT(snk)}$	VIOUT shorted to VCC	–	3	–	mA
Zero Current Reference Voltage	V_{ZCR}	T_A = 25°C	–	2.5	–	V
Zero Current Reference Offset Voltage	$V_{ZCR(ofs)}$	T_A = 25°C	–10	±3	10	mV
		T_A = 25°C to 150°C	–25	±10	25	mV
		T_A = –40°C to 25°C	–25	±10	25	mV
Reference Source Current	$I_{VZCR(src)}$	VZCR shorted to GND	–	2	–	mA
Reference Sink Current	$I_{VZCR(snk)}$	VZCR shorted to VCC	–	14	–	mA

[1] Device may be operated at higher primary current levels, I_P, ambient temperatures, T_A, and internal leadframe temperatures, provided the Maximum Junction Temperature, T_J(max), is not exceeded.

[2] The sensor IC will continue to respond to current beyond the range of IP until the high or low saturation voltage; however, the nonlinearity in this region will be worse than through the rest of the measurement range.

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xLLCTR-40AU PERFORMANCE CHARACTERISTICS: Valid over full range of T_A, V_{CC} = 5 V, C_{BYPASS} = 0.1 µF, unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
NOMINAL PERFORMANCE						
Current Sensing Range	I_{PR}		0	–	40	A
Sensitivity	Sens		–	100	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	I_P = 0 A, T_A = 25°C	–	0.5	–	V
ACCURACY PERFORMANCE						
Total Output Error [2]	E_{TOT}	$I_P = I_{P(MAX)}$; T_A = 25°C to 150°C	–5	±2	5	%
		$I_P = I_{P(MAX)}$; T_A = –40°C to 25°C	–7	±3.5	7	%
Sensitivity Error	E_{sens}	$I_P = I_{P(MAX)}$; T_A = 25°C to 150°C	–3.5	±1.5	3.5	%
		$I_P = I_{P(MAX)}$; T_A = –40°C to 25°C	–6	±3	6	%
Offset Voltage	V_{OE}	I_P = 0 A; T_A = 25°C to 150°C	–75	±20	75	mV
		I_P = 0 A; T_A = –40°C to 25°C	–100	±30	100	mV
Nonlinearity	E_{LIN}	Through the full range of I_P; T_A = 25°C to 150°C	–2	±0.75	2	%
		Through the full range of I_P; T_A = –40°C to 25°C	–3.5	±2.5	3.5	%
LIFETIME DRIFT CHARACTERISTICS [3]						
Total Output Error Including Lifetime Drift	E_{tot_drift}	T_A = 25°C to 150°C	–8.9	±4.8	8.9	%
		T_A = –40°C to 25°C	–8.9	±5.8	8.9	%
Sensitivity Error Including Lifetime Drift	E_{sens_drift}	T_A = 25°C to 150°C	–4.6	±2.8	4.6	%
		T_A = –40°C to 25°C	–4.6	±3.8	4.6	%
Offset Voltage Including Lifetime Drift	V_{off_drift}	T_A = 25°C to 150°C	–165	±83	165	mV
		T_A = –40°C to 25°C	–165	±83	165	mV

[1] Typical values with ± are 3 sigma values.
[2] Percentage of I_P.
[3] Min/max limits based on AEC-Q100 Grade 1 Qualification results.

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Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
NOMINAL PERFORMANCE						
Current Sensing Range	I_{PR}		–50	–	50	A
Sensitivity	Sens		–	40	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	I_P = 0 A, T_A = 25°C	–	2.5	–	V
ACCURACY PERFORMANCE						
Total Output Error [2]	E_{TOT}	$I_P = I_{P(MAX)}$; T_A = 25°C to 150°C	–4	±2	4	%
		$I_P = I_{P(MAX)}$; T_A = –40°C to 25°C	–7	±3.5	7	%
Sensitivity Error	E_{sens}	$I_P = I_{P(MAX)}$; T_A = 25°C to 150°C	–3	±1.5	3	%
		$I_P = I_{P(MAX)}$; T_A = –40°C to 25°C	–6	±3	6	%
Offset Voltage	V_{OE}	I_P = 0 A; T_A = 25°C to 150°C	–40	±20	40	mV
		I_P = 0 A; T_A = –40°C to 25°C	–60	±30	60	mV
Nonlinearity	E_{LIN}	Through the full range of I_P; T_A = 25°C to 150°C	–2	±0.75	2	%
		Through the full range of I_P; T_A = –40°C to 25°C	–3.5	±2.5	3.5	%
LIFETIME DRIFT CHARACTERISTICS [3]						
Total Output Error Including Lifetime Drift	E_{tot_drift}	T_A = 25°C to 150°C	–8.9	±4.8	8.9	%
		T_A = –40°C to 25°C	–8.9	±5.8	8.9	%
Sensitivity Error Including Lifetime Drift	E_{sens_drift}	T_A = 25°C to 150°C	–4.6	±2.8	4.6	%
		T_A = –40°C to 25°C	–4.6	±3.8	4.6	%
Offset Voltage Including Lifetime Drift	V_{off_drift}	T_A = 25°C to 150°C	–165	±83	165	mV
		T_A = –40°C to 25°C	–165	±83	165	mV

[1] Typical values with ± are 3 sigma values.
[2] Percentage of I_P.
[3] Min/max limits based on AEC-Q100 Grade 1 Qualification results.

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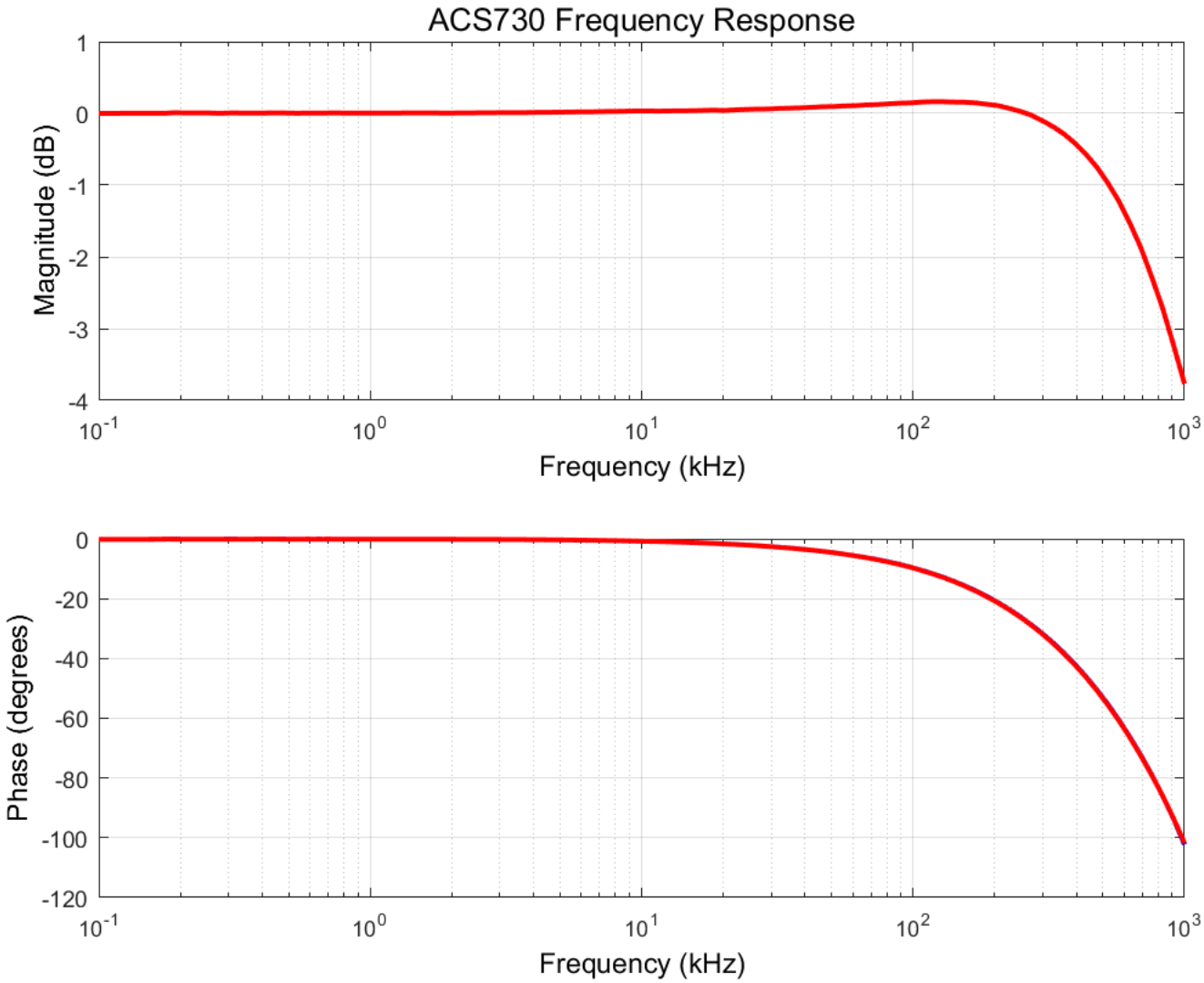
CHARACTERISTIC PERFORMANCE



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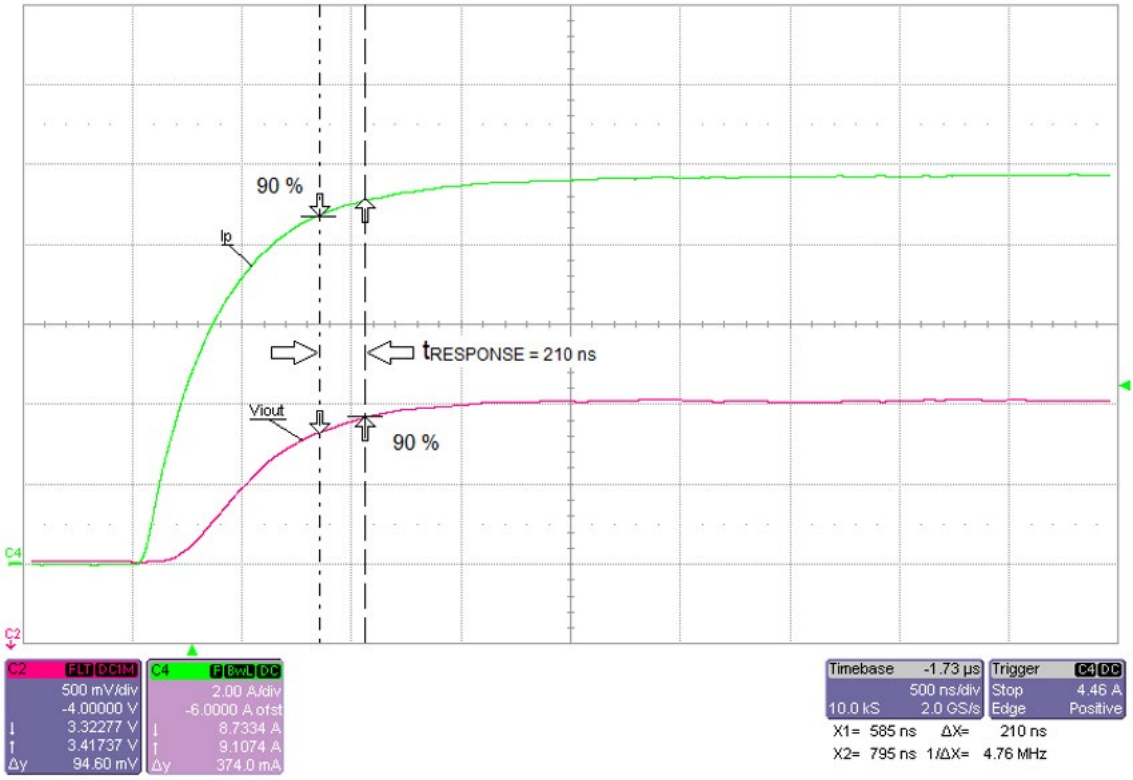
Response Time ($t_{RESPONSE}$)
10 A input signal (I_P) with rise time < 1 µs
Sensitivity = 100 mV/A, C_{BYPASS} = 0.1 µF, C_L = 470 pF, V_{ZCR} = 1 nF



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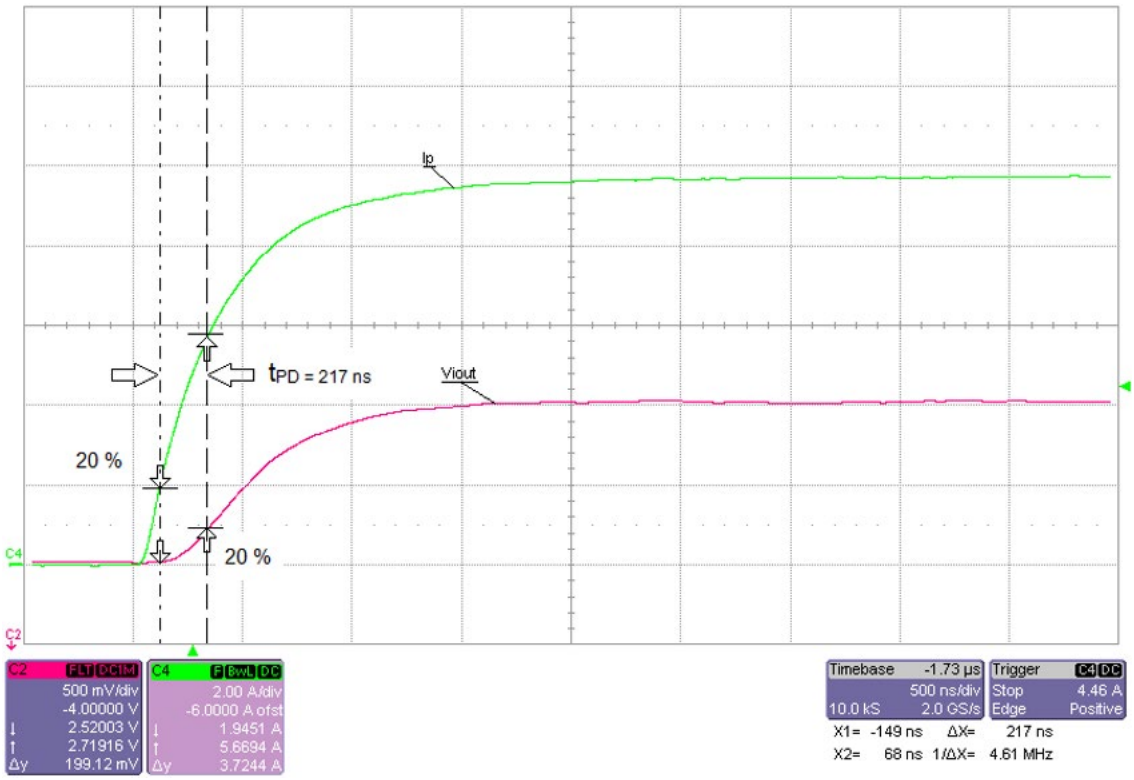
Propagation Delay (t_{PD})

10 A input signal (I_P) with rise time < 1 µs

Sensitivity = 100 mV/A, C_{BYPASS} = 0.1 µF, C_L = 470 pF, V_{ZCR} = 1 nF



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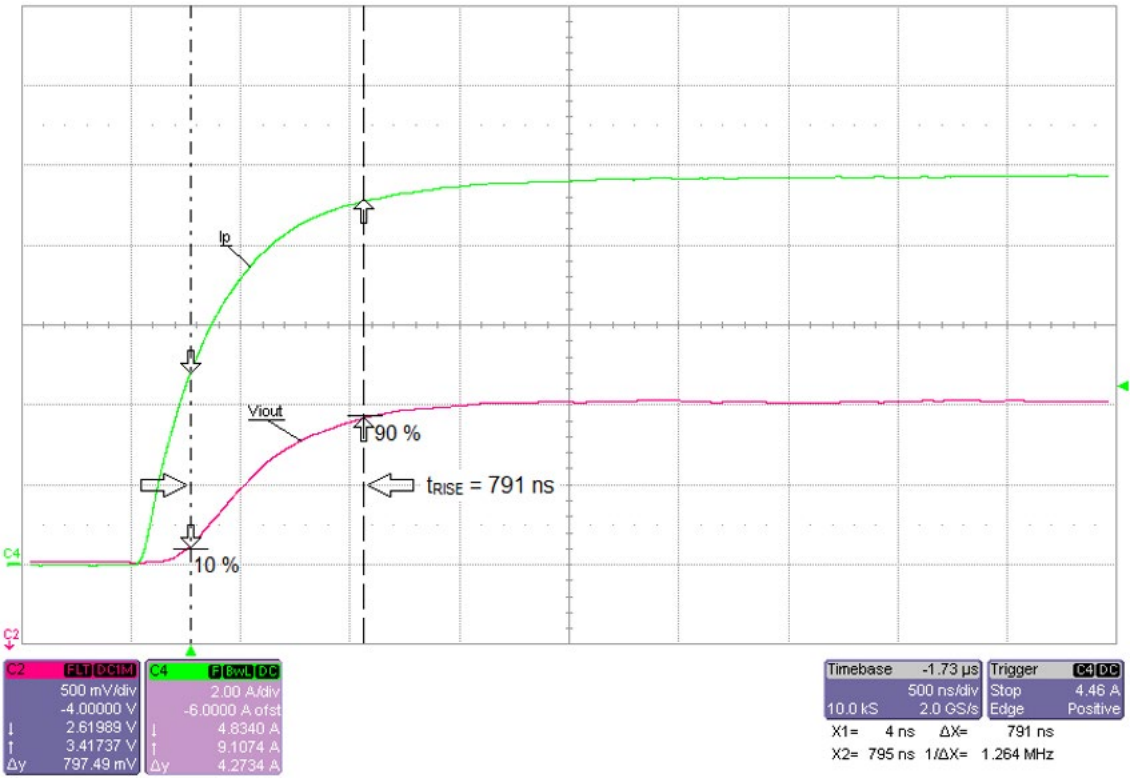
Rise Time (t_r)

10 A input signal (I_P) with rise time < 1 µs

Sensitivity = 100 mV/A, C_{BYPASS} = 0.1 µF, C_L = 470 pF, V_{ZCR} = 1 nF



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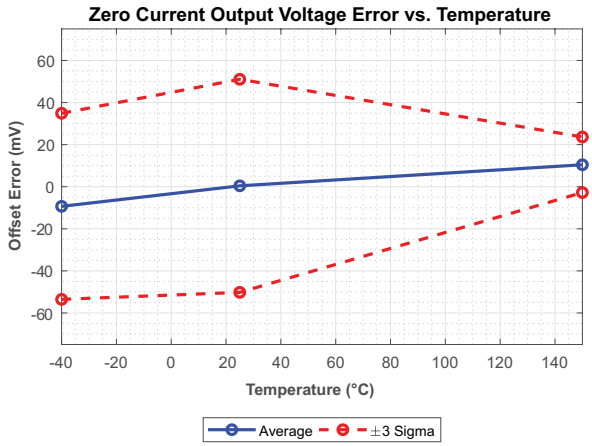
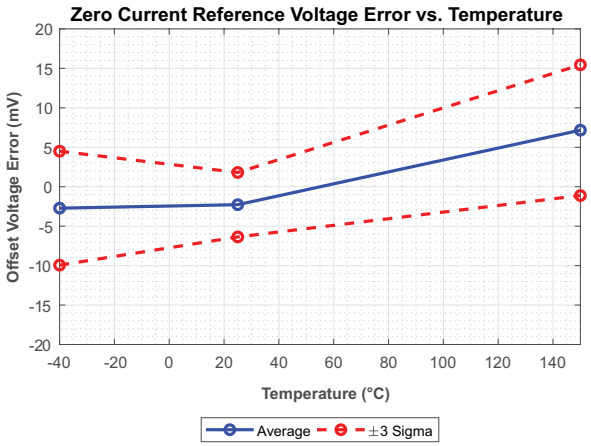
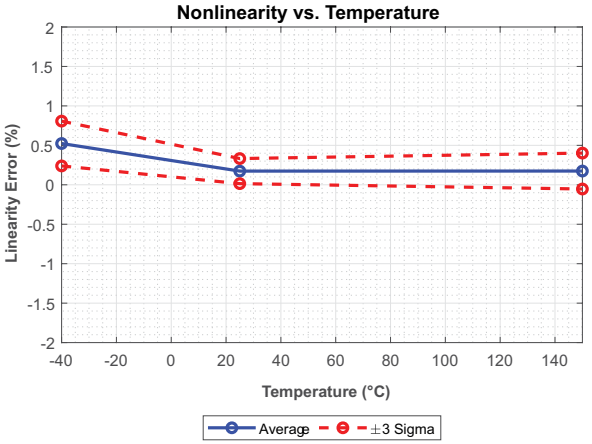
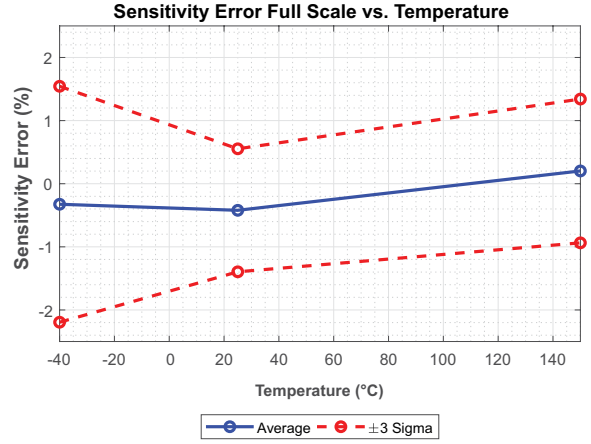
CHARACTERISTIC PERFORMANCE

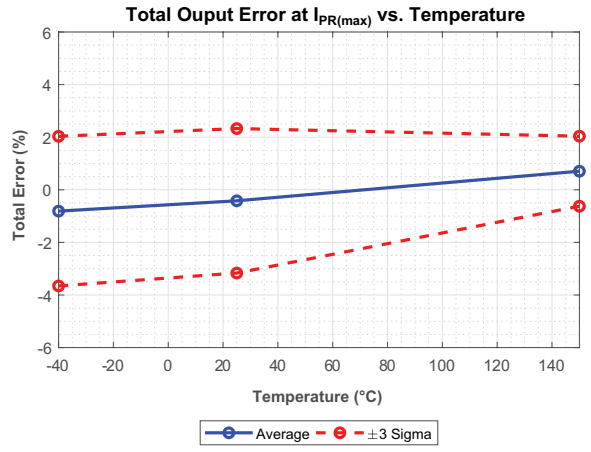
xLLCTR-20AB Key Parameters











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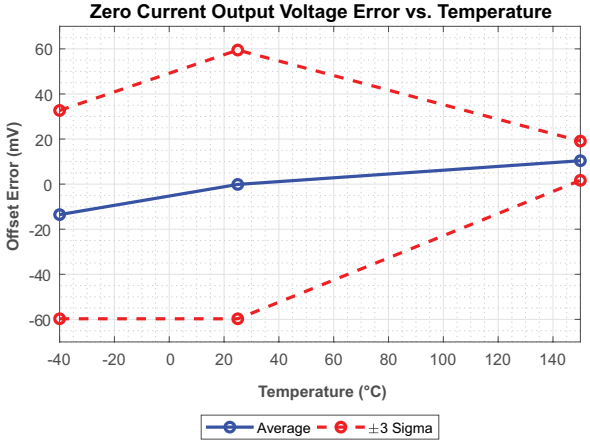
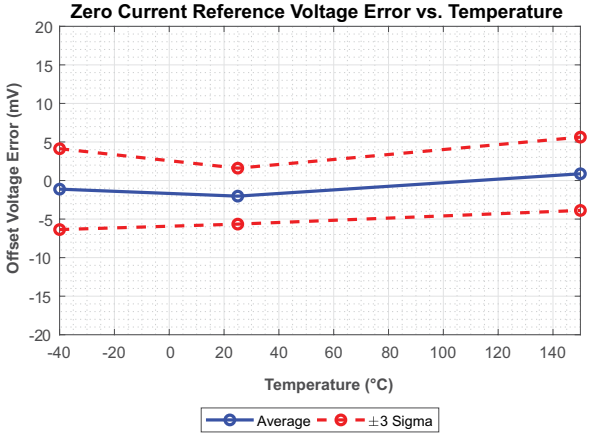
xLLCTR-40AU Key Parameters

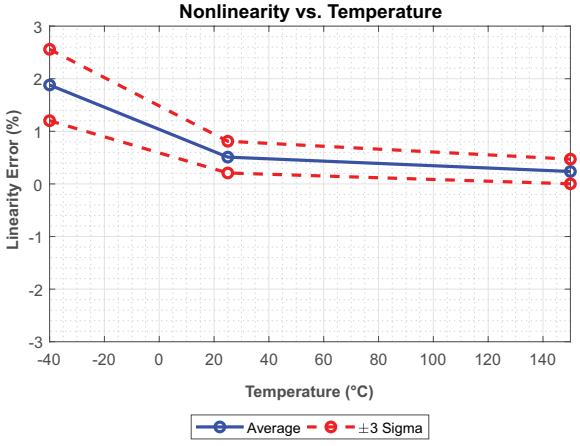
Zero Current Output Voltage Error vs. Temperature



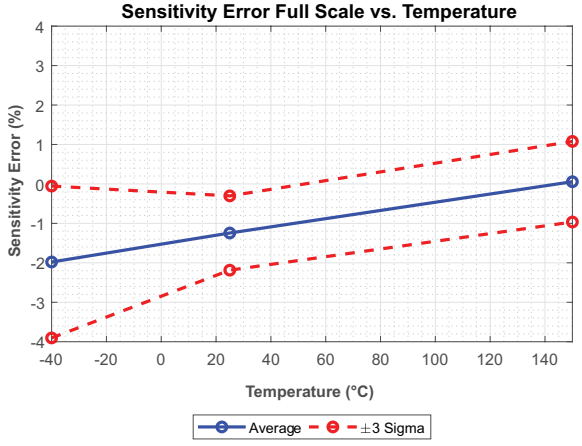
Zero Current Reference Voltage Error vs. Temperature



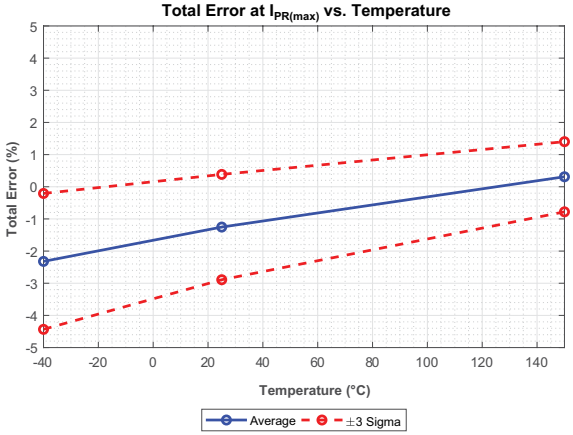
Nonlinearity vs. Temperature



Sensitivity Error Full Scale vs. Temperature



Total Error at $I_{PR(max)}$ vs. Temperature



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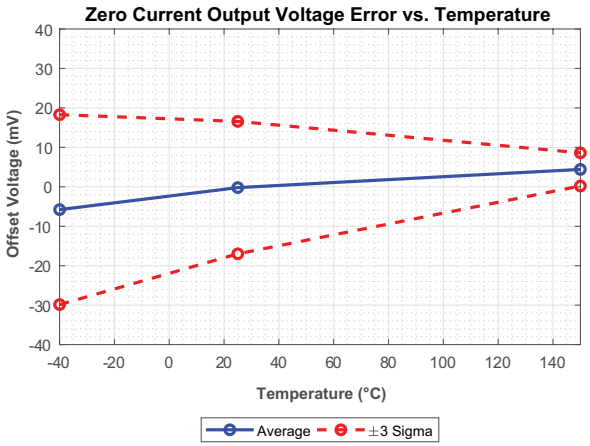
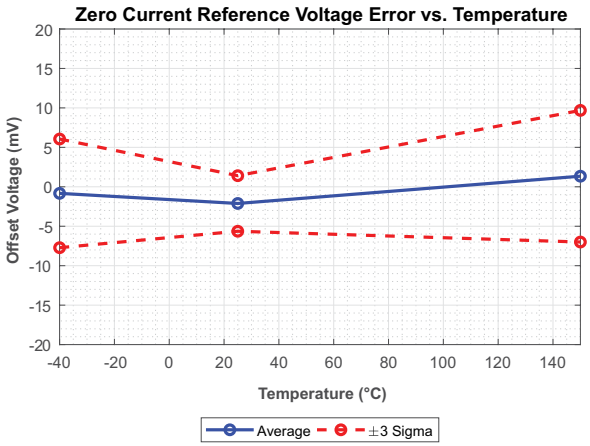
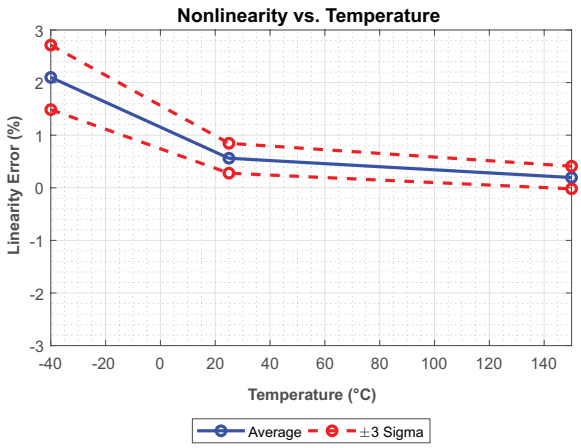
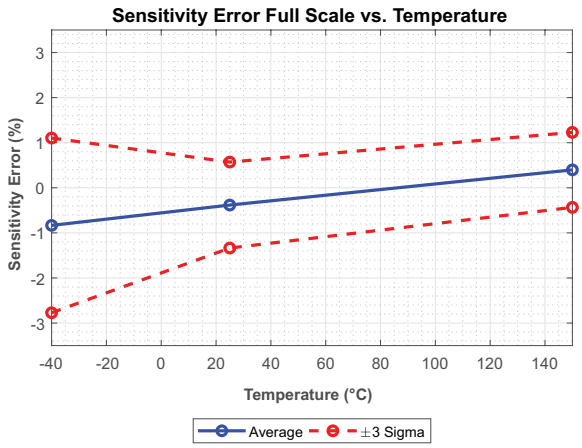
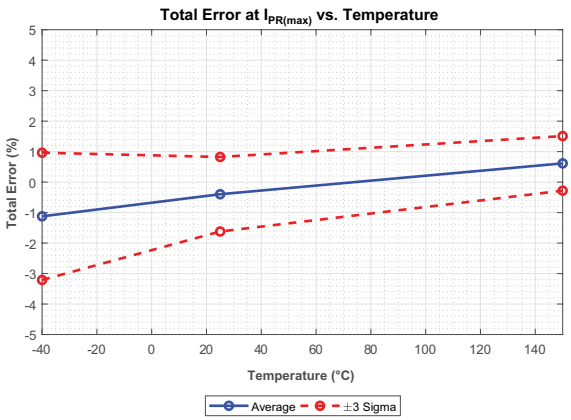
xLLCTR-50AB Key Parameters











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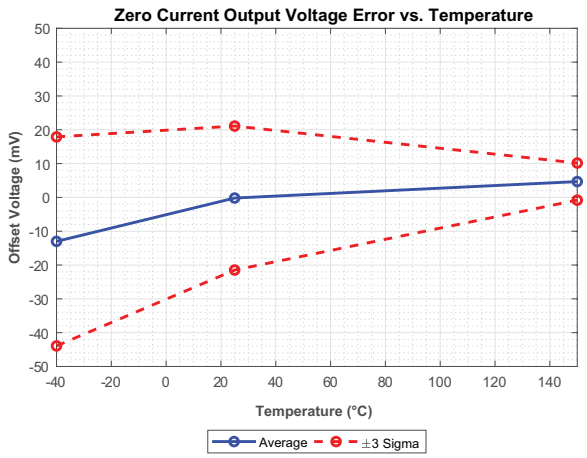
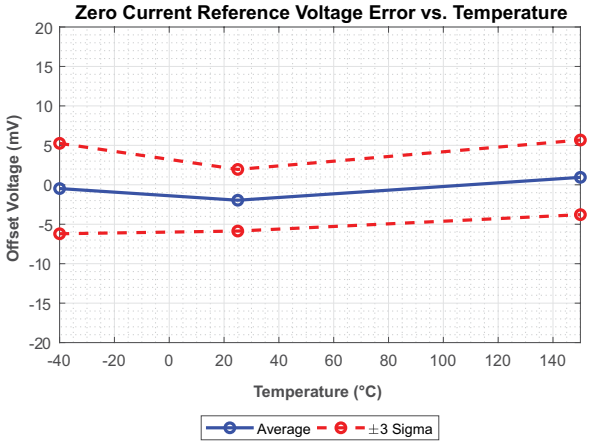
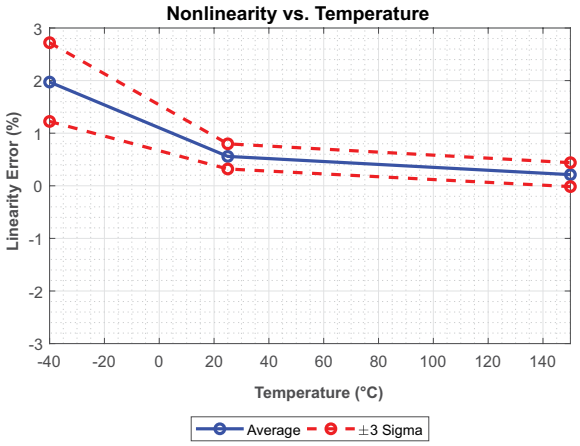
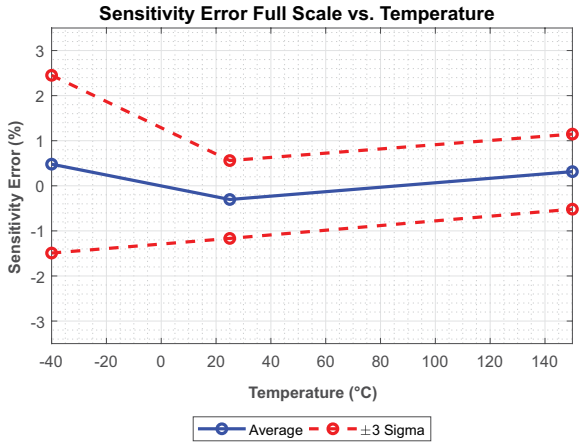
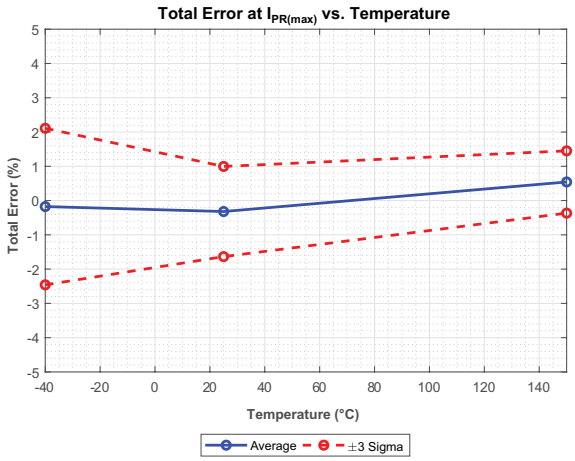
xLLCTR-80AU Key Parameters











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DEFINITIONS OF ACCURACY CHARACTERISTICS

Sensitivity (Sens). The change in sensor IC output in response to a 1 A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) (1 G = 0.1 mT) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the full-scale current of the device.

Nonlinearity (E_{LIN}). The nonlinearity is a measure of how linear the output of the sensor IC is over the full current measurement range. The nonlinearity is calculated as:

$$E_{LIN} = \left\{ 1 - \frac{V_{IOUT}(I_R(max)) - V_{IOUT(Q)}}{2 \bullet V_{IOUT}(I_R(max)/2) - V_{IOUT(Q)}} \right\} \bullet 100(\%)$$

Zero Current Output Voltage ($V_{IOUT(Q)}$). The output of the sensor when the primary current is zero. For a unipolar supply voltage, it nominally remains at 2.5 V for a bidirectional device. Variation in $V_{IOUT(Q)}$ can be attributed to the resolution of the Allegro linear IC quiescent voltage trim and thermal drift.

Offset Voltage (V_{OE}). The deviation of the device output from its ideal quiescent value of 2.5 V due to nonmagnetic causes. To convert this voltage to amperes, divide by the device sensitivity, Sens.

Total Output Error (E_{TOT}). The difference between the current measurement from the sensor IC and the actual current (I_P), relative to the actual current. This is equivalent to the difference between the ideal output voltage and the actual output voltage, divided by the ideal sensitivity, relative to the current flowing through the primary conduction path:

$$E_{TOT}(I_P) = \frac{V_{IOUT_IDEAL}(I_P) - V_{IOUT}(I_P)}{Sens_{IDEAL} \times I_P} \bullet 100\ (\%)$$

The Total Output Error incorporates all sources of error and is a function of I_P. At relatively high currents, E_{TOT} will be mostly due to sensitivity error, and at relatively low currents, E_{TOT} will be mostly due to Offset Voltage (V_{OE}). In fact, at $I_P = 0$, E_{TOT} approaches infinity due to the offset. This is illustrated in Figure 1 and Figure 2. Figure 1 shows a distribution of output voltages versus I_P at 25°C and across temperature. Figure 2 shows the corresponding E_{TOT} versus I_P.

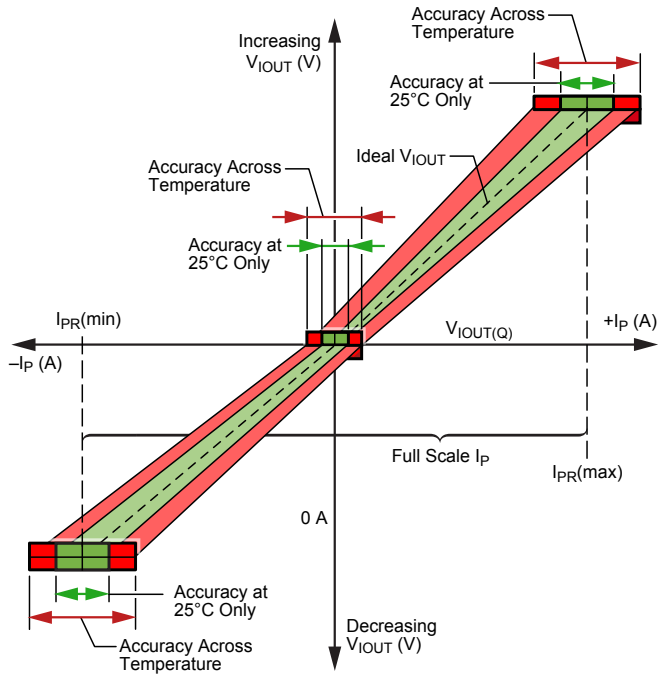

Figure 1: Output Voltage versus Sensed Current

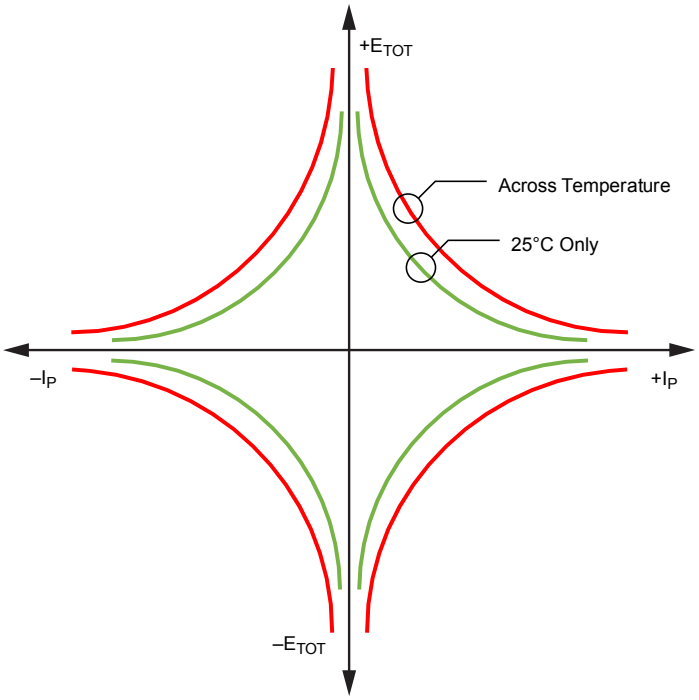

Figure 2: Total Output Error versus Sensed Current

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Power Supply Rejection Ratio (PSRR). The ratio of the change on VIOUT or VZCR to a change in V_{CC} in dB.

$$PSRR = 20 \log_{10} \left(\left| \frac{\Delta V_{CC}}{\Delta V_{IOUT}} \right| \right)$$

Sensitivity Power Supply Rejection Ratio (PSRR). The ratio of the percent change in sensitivity from the sensitivity at 5 V to the percent change in V_{CC} in dB.

$$SPSRR\ (V_{CC}) = 20 \log_{10} \left(\left| \frac{Sens_{VCCN} \times (V_{CC} - 5\ \text{V})}{[Sens_{VCC} - Sens_{5V}] \times 5\ \text{V}} \right| \right)$$

An SPSRR value of 15 dB means that a ten percent change in V_{CC} (going from 5 to 5.5 V, for example) results in around a 1.75 percent change in sensitivity.

Offset Power Supply Rejection Ratio (OPSRR). The ratio of the change in offset to a change in V_{CC} in dB.

$$OPSRR = 20 \log_{10} \left(\left| \frac{\Delta V_{CC}}{\Delta V_{OE}} \right| \right)$$

An OPSRR value of 30 dB means that a 500 mV change in V_{CC} (going from 5 to 5.5 V, for example) results in around 15 mV of change in the offset.

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APPLICATION INFORMATION

Impact of External Magnetic Fields

The ACS730 works by sensing the magnetic field created by the current flowing through the package. However, the sensor cannot differentiate between fields created by the current flow and external magnetic fields. This means that external magnetic fields can cause errors in the output of the sensor. Magnetic fields which are perpendicular to the surface of the package affect the output of the sensor, as it only senses fields in that one plane. The error in Amperes can be quantified as:

$$Error\ (B) = \frac{B}{MCF}$$

where B is the strength of the external field perpendicular to the surface of the package in gauss (G), and MCF is the magnetic coupling factor in gauss/amperes (G/A). Then, multiplying by the sensitivity of the part (Sens) gives the error in mV seen at the output.

For example, an external field of 1 gauss will result in around 0.1 A of error. If the ACS730LLCTR-20AB, which has a nominal sensitivity of 100 mV/A, is being used, that equates to 10 mV of error on the output of the sensor.

External Field (Gauss)	Error (A)	Error (mV)		
		20B	40B	50B
0.5	0.05	5	2.5	2
1	0.1	10	5	4
2	0.2	20	10	8

Estimating Total Error vs. Sensed Current

The Performance Characteristics tables give distribution values (±3 sigma) for Total Error at I_P(max) and I_P(half); however, one often wants to know what error to expect at a particular current. This can be estimated by using the distribution data for the components of Total Error, Sensitivity Error, and Offset Voltage. The ±3 sigma value for Total Error (E_{TOT}) as a function of the sensed current (I_P) is estimated as:

$$E_{TOT}(I_P) = \sqrt{E_{SENS}{}^2 + \left(\frac{100 \times V_{OE}}{Sens \times I_P}\right)^2}$$

Here, E_{SENS} and V_{OE} are the ±3 sigma values for those error terms. If there is an average offset voltage, then the average Total Error is estimated as:

$$E_{TOT_{AVG}}(I_P) = E_{SENS_{AVG}} + \frac{100 \times V_{OE_{AVG}}}{Sens \times I_P}$$

The resulting total error will be a sum of E_{TOT} and E_{TOT_AVG}. Using these equations and the 3 sigma distributions for Sensitivity Error and Offset Voltage, the Total Error versus sensed current (I_P) is below for the ACS730LLCTR-20AB. As expected, as the sensed current (I_P) approaches zero, the error in percent goes towards infinity due to division by zero (refer to Figure 3).

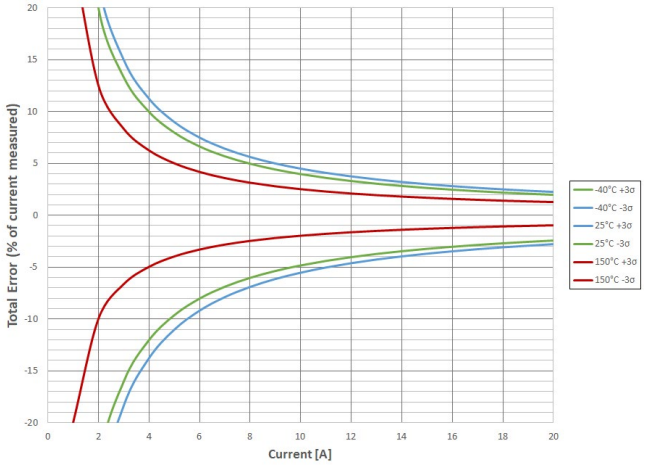

Figure 3: Predicted Total Error as a Function of the Sensed Current for the ACS730LLCTR-20AB

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Thermal Rise vs. Primary Current

Self-heating due to the flow of current should be considered during the design of any current sensing system. The sensor, printed circuit board (PCB), and contacts to the PCB will generate heat as current moves through the system.

The thermal response is highly dependent on PCB layout, copper thickness, cooling techniques, and the profile of the injected current. The current profile includes peak current, current "on-time", and duty cycle. While the data presented in this section was collected with direct current (DC), these numbers may be used to approximate thermal response for both AC signals and current pulses.

The plot in Figure 4 shows the measured rise in steady-state die temperature of the ACS730 versus continuous current at an ambient temperature, T_A, of 25 °C. The thermal offset curves may be directly applied to other values of T_A. Conversely, Figure 5 shows the maximum continuous current at a given T_A. Surges beyond the maximum current listed in Figure 5 are allowed given the maximum junction temperature, $T_{J(MAX)}$ (165°C), is not exceeded.

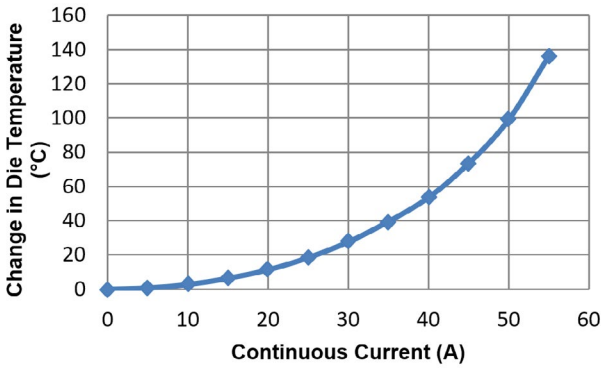

Figure 4: Self Heating in the LC Package Due to Current Flow

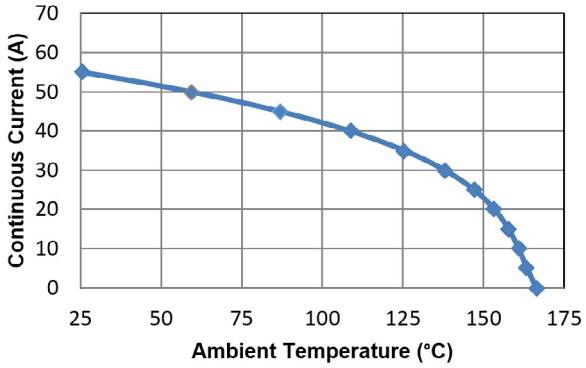

Figure 5: Maximum Continuous Current at a Given T_A

The thermal capacity of the ACS730 should be verified by the end user in the application's specific conditions. The maximum junction temperature, $T_{J(MAX)}$ (165°C), should not be exceeded. Further information on this application testing is available in the DC and Transient Current Capability application note on the Allegro website.

ASEK730 Evaluation Board Layout

Thermal data shown in Figure 4 was collected using the ASEK730 Evaluation Board (TED-85-0739-003). This board includes 1500 mm^2 of 4 oz. copper (0.1388 mm) connected to pins 1 and 2, and to pins 3 and 4, with thermal vias connecting the layers. Top and bottom layers of the PCB are shown below in Figure 6.

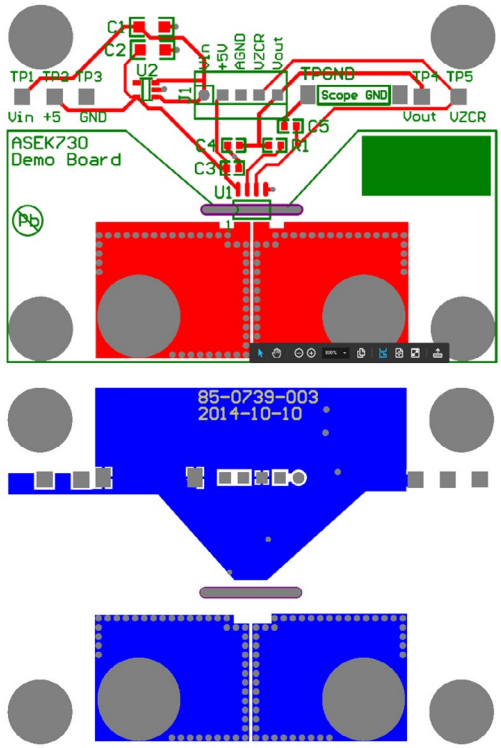

Figure 6: Top and Bottom Layers for ASEK730 Evaluation Board

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DEFINITIONS OF DYNAMIC RESPONSE CHARACTERISTICS

Power-On Time (t_{PO}). When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field. Power-On Time, t_{PO} , is defined as the time it takes for the output voltage to settle within ±10% of its steady-state value under an applied magnetic field, after the power supply has reached its minimum specified operating voltage, V_{CC}(min), as shown in the chart at right.

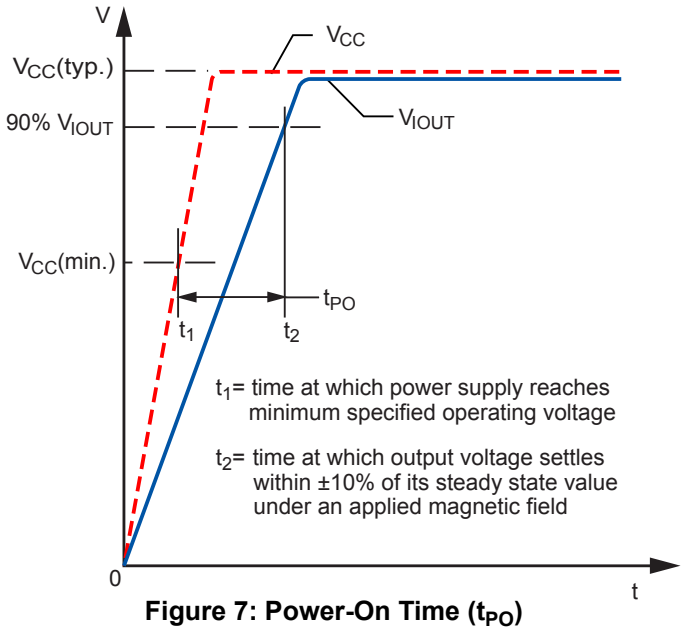

Figure 7: Power-On Time (t_{PO})

Rise Time (t_r). The time interval between a) when the sensor reaches 10% of its full-scale value, and b) when it reaches 90% of its full scale value.

Propagation Delay (t_{pd}). The time interval between a) when the sensed input current reaches 20% of its full-scale value, and b) when the sensor output reaches 20% of its full-scale value.

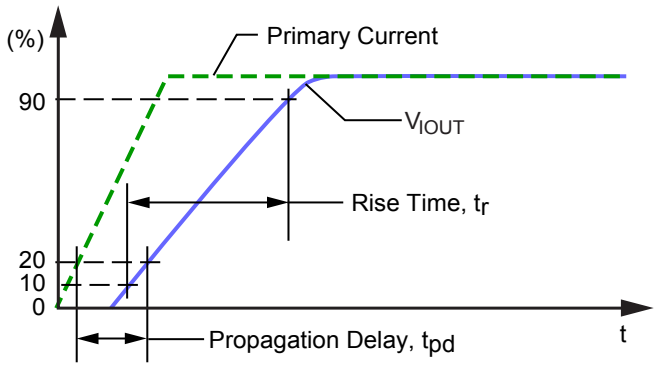

Figure 8: Rise Time (t_r) and Propagation Delay (t_{pd})

Response Time ($t_{RESPONSE}$). The time interval between a) when the sensed input current reaches 90% of its final value, and b) when the sensor output reaches 90% of its full-scale value.

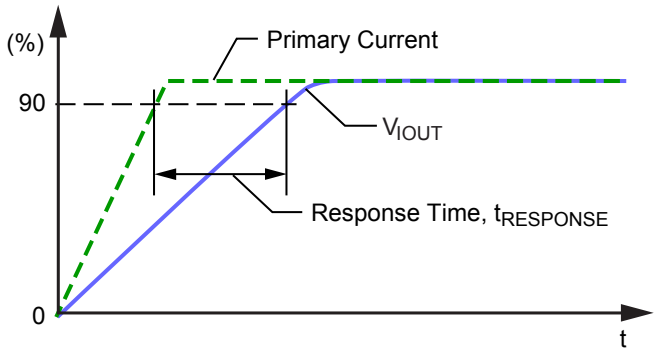

Figure 9: Response Time ($t_{RESPONSE}$)

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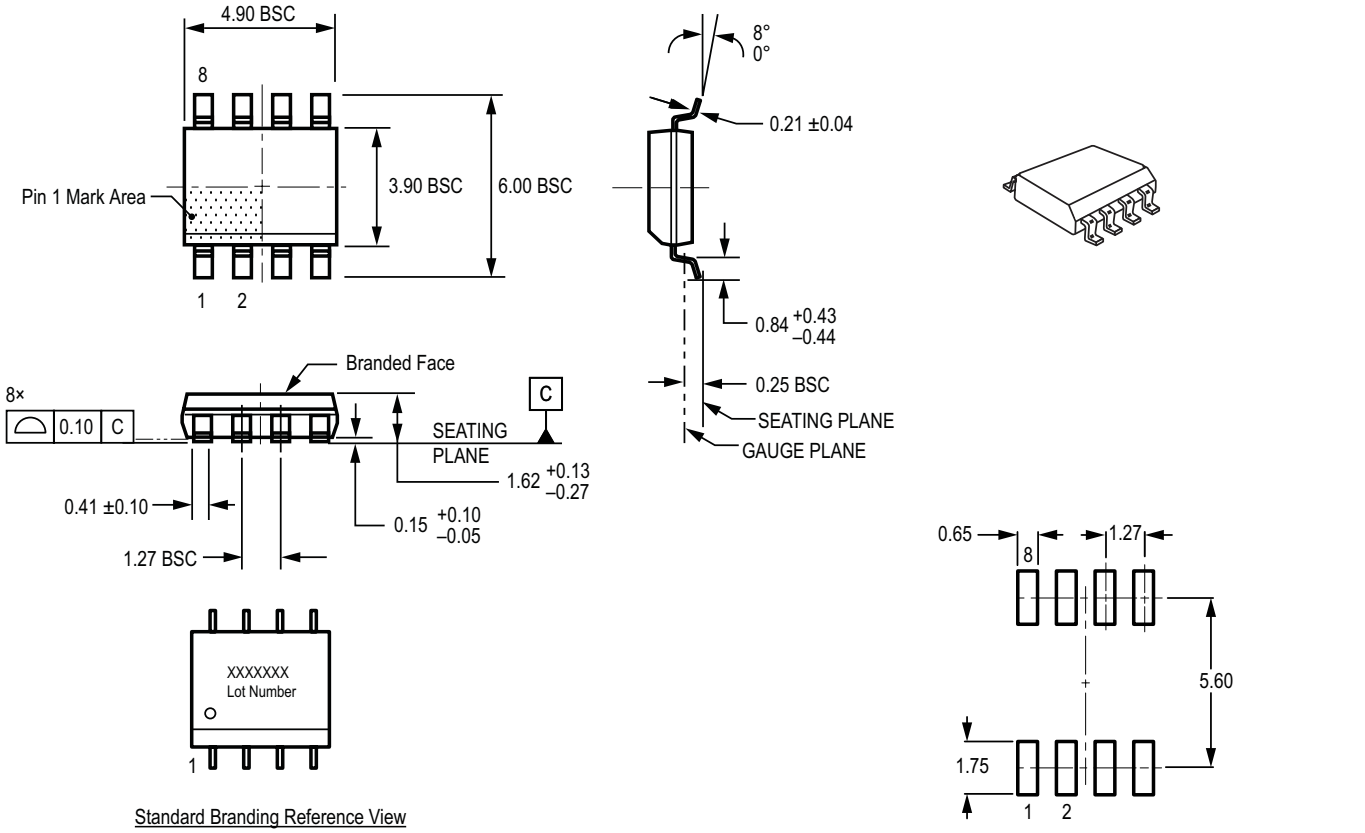
PACKAGE OUTLINE DRAWING

For Reference Only; not for tooling use
(reference Allegro DWG-0000385, Rev. 2 or JEDEC MS-012AA)
Dimensions in millimeters – Not to scale
Dimensions exclusive of mold flash, gate burrs, and dambar protrusions
Exact case and lead configuration at supplier discretion within limits shown



Standard Branding Reference View

Lines 1, 2 = 8 characters

Line 1: Part Number
Line 2: First 8 Characters of Assembly Lot Number

Belly Brand: Country of Origin, Lot Number

Branding scale and appearance at supplier discretion

PCB Layout Reference View
Reference land pattern layout
(reference IPC7351 SOIC127P600X175-8M);
all pads a minimum of 0.20 mm from all adjacent pads;
adjust as necessary to meet application process
requirements and PCB layout tolerances.

Figure 10: Package LC, 8-Pin SOICN

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Revision History

Number	Date	Description
–	January 24, 2019	Initial release
1	May 31, 2019	Updated TUV certificate mark
2	August 21, 2019	Added Maximum Continuous Current to Absolute Maximum Ratings table (page 2), ESD ratings table (page 2), and thermal data section (page 21)
3	May 17, 2022	Added “-S” lead-free part variants; removed non-production part variants; removed ESD Ratings table; removed Maximum Continuous Current; updated package drawing (page 20); minor editorial updates
4	June 14, 2024	Added “-40AU” part variant (page 2, 6); updated Isolation Characteristics Table (page 3)
5	March 12, 2025	Changed product status to not for new design (cover sheet), removed thermal characteristics table (page 3), and removed references to availability of ASEK Gerber files on the web site (page 19)
6	August 28, 2025	Updated recommended substitutions (cover sheet)

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