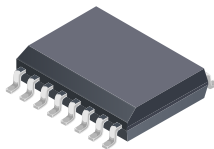


## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

### FEATURES AND BENEFITS

- AEC-Q100 automotive qualified
- High bandwidth, 1 MHz analog output
- Differential Hall sensing rejects common-mode fields
- High-isolation SOIC-16 wide-body package provides galvanic isolation for high-voltage applications
- Integrated shield virtually eliminates capacitive coupling from current conductor to die, greatly suppressing output noise due to high dV/dt transients
- Industry-leading noise performance with greatly improved bandwidth through proprietary amplifier and filter design
- UL 62368-1 (ed. 3) certified
  - Dielectric strength voltage = 4242 V<sub>RMS</sub>
  - Basic isolation working voltage = 1000 V<sub>RMS</sub>
- Fast and externally configurable overcurrent fault detection
- 1 mΩ primary conductor resistance for low power loss and high inrush current withstand capability
- Options for 3.3 V and 5 V single-supply operation
- Output voltage proportional to AC and DC current
- Factory-trimmed sensitivity and quiescent output voltage for improved accuracy
- Nearly zero magnetic hysteresis
- Ratiometric output from supply voltage

### PACKAGE: 16-Pin SOICW (suffix LA)



Not to scale



### DESCRIPTION

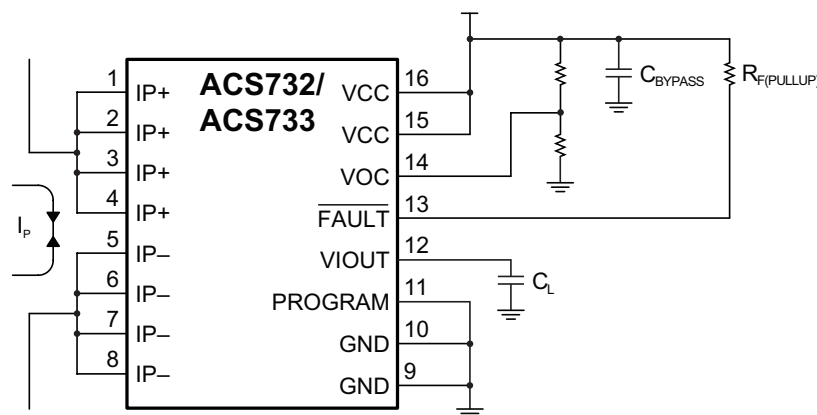
The ACS732 and ACS733 are a new generation of high-bandwidth current sensor ICs from Allegro. These devices provide a compact, fast, and accurate solution for measuring high-frequency currents in DC-to-DC converters and other switching power applications. The ACS732 and ACS733 offer high-isolation, high-bandwidth Hall-effect-based current sensing with user-configurable overcurrent fault detection. These features make them ideally suited for high-frequency transformer and current transformer replacement in applications running at high voltages.

The ACS732 and ACS733 are suitable for all markets, including automotive, industrial, commercial, and communications systems. They may be used in motor control, load detection and management, switch-mode power supplies, and overcurrent fault protection applications.

The wide-body SOIC-16 package allows for easy implementation. Applied current flowing through the copper conduction path generates a magnetic field that is sensed by the IC and converted to a proportional voltage. Current is sensed differentially in order to reject external common-mode fields. Device accuracy is optimized through the close proximity of the magnetic field to the Hall transducers. A precise, proportional voltage is provided by the Hall IC, which is factory-programmed after packaging for high accuracy. The fully integrated package has an internal copper conductive path with a typical resistance of 1 mΩ, providing low power loss.

The current-carrying pins (pins 1 through 8) are electrically isolated from the sensor leads (pins 9 through 16). This allows the devices to be used in high-side current sensing applications without the use of high-side differential amplifiers or other costly isolation techniques.

*Continued on next page...*



ACS732/ACS733 outputs an analog signal, V<sub>IOUT</sub>, that changes proportionally with the bidirectional AC or DC primary sensed current, I<sub>p</sub>, within the specified measurement range.

The overcurrent threshold may be set with a resistor divider tied to the V<sub>OC</sub> pin.

Figure 1: Typical Application Circuit

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

### DESCRIPTION (continued)

The ACS732 and ACS733 are provided in a small, low-profile, surface-mount SOIC-16 wide-body package. The leadframe is plated with 100% matte tin, which is compatible with standard lead (Pb)

free printed circuit board assembly processes. Internally, the device is lead-free. These devices are fully calibrated prior to shipment from the Allegro factory.

### SELECTION GUIDE

Part Number	Optimized Range, $I_P$ (A)	Sensitivity, <sup>[1]</sup> Sens(Typ) (mV/A)	Nominal Supply Voltage, $V_{CC}$ , (V)	$T_A$ (°C)	Packing <sup>[2]</sup>
ACS732KLATR-20AB-T	±20	100	5.0	-40 to 125	Tape and reel, 1000 pieces per reel
ACS732KLATR-40AB-T	±40	50			
ACS732KLATR-65AB-T	±65	30			
ACS732KLATR-65AU-T	65	60			
ACS732KLATR-75AB-T	±75	26.6			
ACS733KLATR-20AB-T	±20	66	3.3		
ACS733KLATR-20AB-T-H <sup>[3]</sup>	±20	66			
ACS733KLATR-40AB-T	±40	33			
ACS733KLATR-40AU-T	40	66			
ACS733KLATR-65AB-T	±65	20			

<sup>[1]</sup> Measured at nominal supply voltage,  $V_{CC}$ .

<sup>[2]</sup> Contact Allegro for additional packing options.

<sup>[3]</sup> -H denotes 100% cold calibration at the Allegro factory for improved accuracy.



# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

### ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Notes	Rating	Units
Supply Voltage	$V_{CC}$		6	V
Reverse Supply Voltage	$V_{RCC}$		-0.1	V
Output Voltage	$V_{IOUT}$		6	V
Reverse Output Voltage	$V_{RIOUT}$		-0.1	V
Fault Output Voltage	$V_{FAULT}$		6	V
Reverse Fault Output Voltage	$V_{RFAULT}$		-0.1	V
Forward $V_{OC}$ Voltage	$V_{VOC}$		6	V
Reverse $V_{OC}$ Voltage	$V_{VOC}$		-0.1	V
Output Current	$I_{OUT}$	Maximum survivable sink or source current on the output	15	mA
Nominal Operating Ambient Temperature	$T_A$	Range K	-40 to 125	°C
Maximum Junction Temperature	$T_J(max)$		165	°C
Storage Temperature	$T_{stg}$		-65 to 170	°C

### ESD RATINGS

Characteristic	Symbol	Test Conditions	Value	Unit
Human Body Model	$V_{HBM}$	Per AEC-Q100	±12	kV
Charged Device Model	$V_{CDM}$	Per AEC-Q100	±1	kV

### LA PACKAGE ISOLATION CHARACTERISTICS

Characteristic	Symbol	Notes	Rating	Unit
Withstand Voltage <sup>[1][2]</sup>	$V_{ISO}$	Agency rated for 60 seconds per UL 62368-1 (edition 3)	4242	$V_{RMS}$
Working Voltage for Basic Isolation <sup>[2]</sup>	$V_{WVBI}$	Maximum approved working voltage for basic (single) isolation according to UL 62368-1 (edition 3)	1414	$V_{PK}$ or $V_{DC}$
			1000	$V_{RMS}$
Working Voltage for Reinforced Isolation <sup>[2]</sup>	$V_{WVRI}$	Maximum approved working voltage for reinforced isolation according to UL 62368-1 (edition 3)	707	$V_{PK}$ or $V_{DC}$
			500	$V_{RMS}$
Surge Voltage	$V_{SURGE}$	1.2/50 $\mu$ s waveform, tested in dielectric fluid to determine the intrinsic surge immunity of the isolation barrier	10000	$V_{PK}$
Impulse Withstand Voltage	$V_{IMPULSE}$	1.2/50 $\mu$ s waveform tested in air	6000	$V_{PK}$
Clearance	$D_{CL}$	Minimum distance through air from IP leads to signal leads	8	mm
Creepage	$D_{CR}$	Minimum distance along package body from IP leads to signal leads	8	mm
Distance Through Insulation	DTI	Minimum internal distance through insulation	64	$\mu$ m
Comparative Tracking Index	CTI	Material Group II	400 to 599	V

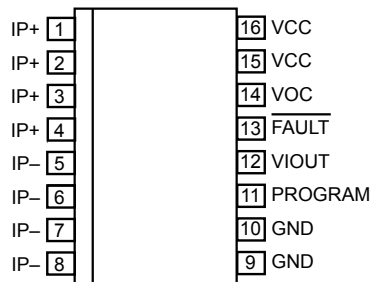
<sup>[1]</sup> Production-tested in accordance with UL 62368-1 (edition 3).

<sup>[2]</sup> Certification pending.

### THERMAL CHARACTERISTICS

Characteristic	Symbol	Notes	Value	Unit
Package Thermal Resistance (Junction to Ambient)	$R_{\theta JA}$	Mounted on the standard MA/LA current sensor evaluation board (ACSEVB-MA16-LA16)	16	°C/W
Package Thermal Characterization (Junction to Top)	$\Psi_{JT}$		-1.7	°C/W
Package Thermal Resistance (Junction to Case)	$R_{\theta JC}$	Simulated per the methods in JESD51-1	10	°C/W
Package Thermal Resistance (Junction to Board)	$R_{\theta JB}$	Simulated per the methods in JESD51-8	8	°C/W

### PINOUT DIAGRAM AND TERMINAL LIST TABLE



**Package LA, 16-Pin  
SOICW Pinout Diagram**

**Terminal List Table**

Number	Name	Description
1,2,3,4	IP+	Positive terminals for current being sensed; fused internally.
5,6,7,8	IP-	Negative terminals for current being sensed; fused internally.
9,10	GND	Device ground terminal.
11	PROGRAM	Programming input pin for factory calibration. Connect to ground for best ESD performance.
12	VIOUT	Analog output signal.
13	FAULT	Overcurrent fault output. Open drain.
14	VOC	Set the overcurrent fault threshold via external resistor divider on this pin.
15,16	VCC	Device power supply terminal.

## FUNCTIONAL BLOCK DIAGRAM

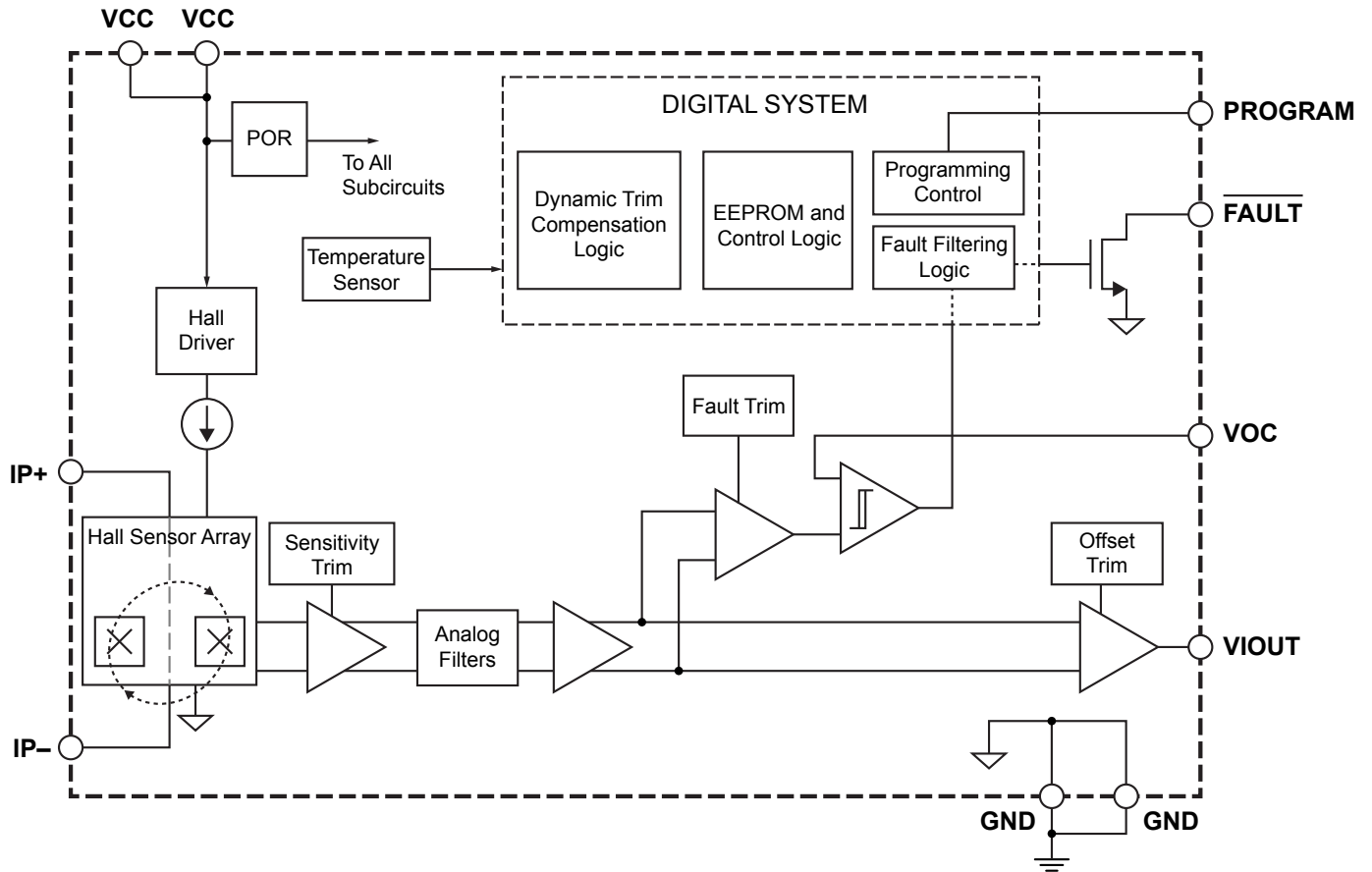


Figure 2: Functional Block Diagram

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**COMMON ELECTRICAL CHARACTERISTICS:** Over full range of  $T_A$ , over supply voltage range  $V_{CC(MIN)}$  through  $V_{CC(MAX)}$  of a sensor variant,  $C_{BYPASS} = 0.1 \mu F$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
Supply Voltage	$V_{CC}$	ACS732	4.75	5.0	5.25	V
		ACS733	3.14	3.3	3.46	V
Supply Current	$I_{CC}$	ACS732; $V_{CC} = 5.0 V$	–	24	35	mA
		ACS733; $V_{CC} = 3.3 V$	–	20	35	mA
Bypass Capacitor [2]	$C_{BYPASS}$	$V_{CC}$ to GND	0.1	–	–	$\mu F$
Output Capacitance Load	$C_L$	$V_{IOUT}$ to GND	–	–	220	pF
Output Resistive Load	$R_L$	$V_{IOUT}$ to GND	50	–	–	k $\Omega$
Output Saturation Voltage	$V_{SAT(HIGH)}$	$V_{CC} = 5.0 V, T_A = 25^\circ C,$ $R_{L(PULLDOWN)} = 50 k\Omega$ to GND	$V_{CC} - 0.3$	–	–	V
		$V_{CC} = 3.3 V, T_A = 25^\circ C,$ $R_{L(PULLDOWN)} = 50 k\Omega$ to GND	$V_{CC} - 0.3$	–	–	V
	$V_{SAT(LOW)}$	$V_{CC} = 5.0 V, T_A = 25^\circ C,$ $R_{L(PULLDOWN)} = 50 k\Omega$ to VCC	–	–	0.3	V
		$V_{CC} = 3.3 V, T_A = 25^\circ C,$ $R_{L(PULLDOWN)} = 50 k\Omega$ to VCC	–	–	0.2	V
Primary Conductor Resistance	$R_{IP}$	$T_A = 25^\circ C$	–	1	–	m $\Omega$
Primary Hall Coupling Factor	$C_{F(P)}$	$T_A = 25^\circ C$	–	10.8	–	G/A
Secondary Hall Coupling Factor	$C_{F(S)}$	$T_A = 25^\circ C$	–	2.4	–	G/A
Hall Plate Sensitivity Matching	$Sens_{match}$	$T_A = 25^\circ C$	–	1	–	%
Power On Delay Time	$t_{POD}$	$T_A = 25^\circ C$	–	180	–	$\mu s$
Internal Bandwidth	BW	Small signal –3 dB; $C_L = 220 pF$	–	1	–	MHz
Rise Time [3]	$t_r$	$T_A = 25^\circ C, C_L = 0.22 nF$	–	0.7	–	$\mu s$
Response Time [3]	$t_{RESPONSE}$	$T_A = 25^\circ C, C_L = 0.22 nF$	–	0.2	–	$\mu s$
Propagation Delay Time [3]	$t_{pd}$	$T_A = 25^\circ C, C_L = 0.22 nF$	–	0.14	–	$\mu s$
Output Slew Rate	SR	$T_A = 25^\circ C, C_L = 0.22 nF$	–	3.2	–	V/ $\mu s$
Zero Current Output Ratiometry Error	$E_{RAT(Q)}$	$T_A = 25^\circ C, V_{CC} = \pm 5\%$ variation of nominal supply voltage	–12	$\pm 10$	12	mV
Sensitivity Ratiometry Error	$E_{RAT(SENS)}$	$T_A = 25^\circ C, V_{CC} = \pm 5\%$ variation of nominal supply voltage	–2	$\pm 1.72$	2	%
Ratiometry Bandwidth	$BW_{RAT}$	$\pm 100 mV$ on $V_{CC}$	–	10	–	kHz
Common Mode Transient Immunity	CMTI	$V_{CM} = 1000 V$ SR = 140 V/ns; $V_{OUT}$ within 100 mV; $R_{L(MIN)}$	–	150	–	ns
		$V_{CM} = 1000 V$ ; $V_{OUT}$ within 100 mV in 300 ns; $R_{L(MIN)}$	100	200	–	V/ns
Noise Density	$I_{ND}$	$V_{CC} = 5.0 V, T_A = 25^\circ C, C_L = 220 pF$ ; input referred	–	55	–	$\mu A/\sqrt{Hz}$
		$V_{CC} = 3.3 V, T_A = 25^\circ C, C_L = 220 pF$ ; input referred	–	80	–	$\mu A/\sqrt{Hz}$

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**COMMON ELECTRICAL CHARACTERISTICS (continued):** Over full range of  $T_A$ , over supply voltage range  $V_{CC(MIN)}$  through  $V_{CC(MAX)}$  of a sensor variant,  $C_{BYPASS} = 0.1 \mu F$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>OVERCURRENT FAULT CHARACTERISTICS</b>						
$\overline{FAULT}$ Response Time [4]	$t_{RESPONSE(F)}$	Time from $I_P > I_{FAULT}$ to when $\overline{FAULT}$ pin is pulled below $V_{\overline{FAULT}}$ ; input current step from 0 to $1.2 \times I_{FAULT}$	0.2	0.5	0.75	$\mu s$
$\overline{FAULT}$ Release Time [4]	$t_{C(F)}$	Time from $I_P$ falling below $I_{FAULT} - I_{HYS}$ to when $V_{\overline{FAULT}}$ is pulled above $V_{\overline{FAULTL}}$ ; 100 pF from $\overline{FAULT}$ to ground	0.1	–	0.45	$\mu s$
$\overline{FAULT}$ Range	$I_{\overline{FAULT}}$	Relative to the full scale of $I_{PR}$ ; set via the VOC pin	$0.5 \times I_{PR}$	–	$2 \times I_{PR}$	A
$\overline{FAULT}$ Output Low Voltage	$V_{\overline{FAULT}}$	In fault condition; $R_{F(PULLUP)} = 10 k\Omega$	–	–	0.4	V
$\overline{FAULT}$ Pull-Up Resistance	$R_{F(PULLUP)}$		10	–	500	k $\Omega$
$\overline{FAULT}$ Leakage Current	$I_{\overline{FAULT}(LEAKAGE)}$		–	$\pm 2$	–	nA
$\overline{FAULT}$ Hysteresis [5]	$I_{HYS}$		–	$0.05 \times I_{PR}$	–	A
$\overline{FAULT}$ Error [6]	$E_{\overline{FAULT}}$	Tested at $V_{VOC} = 0.2 \times V_{CC}$ ( $I_{\overline{FAULT}}$ threshold = $100\% \times I_{PR}$ )	–	$\pm 5$	–	%
$V_{OC}$ Input Range	$V_{VOC}$		$0.1 \times V_{CC}$	–	$0.4 \times V_{CC}$	V
$V_{OC}$ Input Current	$I_{VOC}$		–	10	100	nA

[1] Typical values are mean  $\pm 3$  sigma values.

[2] Use of a bypass capacitor is required to increase output stability.

[3] See definitions of Dynamic Response Characteristics section of this datasheet.

[4] Guaranteed by design.

[5] After  $I_P$  goes above  $I_{\overline{FAULT}}$ , tripping the internal comparator,  $I_P$  must fall below  $I_{\overline{FAULT}} - I_{HYS}$ , before the internal comparator will reset.

[6] Fault error is defined as the value at which a fault is reported relative to the desired threshold for  $I_{\overline{FAULT}}$ .

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS732KLATR-20AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-20	-	20	A
Sensitivity	Sens		-	100	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-3	$\pm 2$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.6$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-7.5	$\pm 4.5$	7.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-1.5	$\pm 0.75$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-1.5	$\pm 1.25$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-3	$\pm 2$	3	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-55	$\pm 30$	55	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 18$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 50$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-120	$\pm 100$	120	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1	$\pm 0.2$	1	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.3$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.5$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 11.5$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 11.5$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 220$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 220$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.



# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS732KLATR-40AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-40	-	40	A
Sensitivity	Sens		-	50	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-2.5	$\pm 2$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.6$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-6.5	$\pm 4.5$	6.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-2	$\pm 0.8$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-2	$\pm 1.2$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-4	$\pm 1.8$	4	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-45	$\pm 40$	45	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 20$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 45$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-95	$\pm 90$	95	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1.2	$\pm 0.8$	1.2	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.33$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 2.5$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 6.3$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 6.3$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 110$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 110$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS732KLATR-65AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-65	-	65	A
Sensitivity	Sens		-	30	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-2.5	$\pm 1$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.6$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-6.5	$\pm 3.4$	6.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-2	$\pm 1.5$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-2	$\pm 0.9$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-4	$\pm 2.7$	4	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-45	$\pm 27$	45	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 8$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 45$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-95	$\pm 58$	95	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-2.4	$\pm 1.7$	2.4	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.6$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 5.8$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 4.7$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 4.7$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 66$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 66$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS732KLATR-65AU PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		0	–	65	A
Sensitivity	Sens		–	60	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Unidirectional, $I_P = 0\text{ A}$	–	$0.1 \times V_{CC}$	–	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	–2.5	$\pm 1.8$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	–3	$\pm 2$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	–	$\pm 2.9$	–	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	–8	$\pm 4.5$	8	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	–1.5	$\pm 0.75$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	–1.5	$\pm 1.25$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	–4	$\pm 2$	4	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	–55	$\pm 30$	55	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	–25	$\pm 18$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	–	$\pm 50$	–	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	–120	$\pm 100$	120	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	–2.5	$\pm 1.8$	2.5	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	–	$\pm 0.7$	–	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	–	$\pm 4.2$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 7.4$	–	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 7.4$	–	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 2.2$	–	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 3.3$	–	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 132$	–	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 132$	–	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS732KLATR-75AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range [2]	$I_{PR}$		-75	-	75	A
Sensitivity	Sens		-	26.6	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [3] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [4]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-2.5	$\pm 1$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.6$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-6.5	$\pm 3.4$	6.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-2	$\pm 1.5$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-2	$\pm 0.9$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-4	$\pm 2.7$	4	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-45	$\pm 27$	45	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 8$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 45$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-95	$\pm 58$	95	mV
Linearity Error [5]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-2.9	$\pm 2.3$	2.9	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 1$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 8.1$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [6] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 4.4$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 4.4$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 59$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 59$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] Devices trimmed at half-scale  $I_P$ . Operating above this limit may result in decreased accuracy.

[3] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[4] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)} / 2$ .

[5] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[6] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733, lifetime drift limits apply immediately after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS733KLATR-20AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-20	-	20	A
Sensitivity	Sens		-	66	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-4.5	$\pm 1.7$	4.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-3	$\pm 1.25$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.8$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-10	$\pm 5$	10	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-1.5	$\pm 1$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-1.5	$\pm 0.8$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-3	$\pm 2$	3	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-55	$\pm 21$	55	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 10$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 35$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-120	$\pm 80$	120	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1	$\pm 0.3$	1	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.4$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.4$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 9.8$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 9.8$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 145$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 145$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733, lifetime drift limits apply immediately after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS733KLATR-20AB-H PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-20	-	20	A
Sensitivity	Sens		-	66	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-4.5	$\pm 1.7$	4.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-3	$\pm 1.25$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.8$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-4.5	$\pm 1.7$	4.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-1.5	$\pm 1$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-1.5	$\pm 0.8$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-1.5	$\pm 1$	1.5	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-55	$\pm 21$	55	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 10$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 35$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-55	$\pm 21$	55	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1	$\pm 0.3$	1	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.4$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.4$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 9.8$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 9.8$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 145$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 145$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733, lifetime drift limits apply immediately after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS733KLATR-40AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-40	-	40	A
Sensitivity	Sens		-	33	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-3	$\pm 1.4$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-2	$\pm 1.25$	2	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.3$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-6.5	$\pm 3$	6.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-1.5	$\pm 1.3$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-2	$\pm 1$	2	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-4.5	$\pm 2.2$	4.5	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-40	$\pm 9$	40	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-40	$\pm 7$	40	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 15$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-75	$\pm 35$	75	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1	$\pm 0.5$	1	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.3$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 1.3$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 4.9$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 4.9$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 73$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 73$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733, lifetime drift limits apply after solder reflow.

# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS733KLATR-40AU PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		0	–	40	A
Sensitivity	Sens		–	66	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Unidirectional, $I_P = 0\text{ A}$	–	$0.1 \times V_{CC}$	–	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	–2.5	$\pm 1.9$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	–2.5	$\pm 1.5$	2.5	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	–	$\pm 3$	–	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	–6.5	$\pm 4$	6.5	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	–1.5	$\pm 0.9$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	–1.5	$\pm 1.5$	1.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	–4	$\pm 1.9$	4	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	–30	$\pm 25$	30	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	–25	$\pm 10$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	–	$\pm 70$	–	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	–110	$\pm 80$	110	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	–1	$\pm 0.6$	1	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	–	$\pm 0.3$	–	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	–	$\pm 1.5$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 8$	–	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 8$	–	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 2.2$	–	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 3.3$	–	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	–	$\pm 145$	–	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	–	$\pm 145$	–	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.



# ACS732 and ACS733

## 1 MHz Bandwidth, Galvanically Isolated Current Sensor IC in SOIC-16 Package

**ACS733KLATR-65AB PERFORMANCE CHARACTERISTICS:** Valid at  $T_A = -40^\circ\text{C}$  to  $125^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ ,  $C_{BYPASS} = 0.1\ \mu\text{F}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-65	-	65	A
Sensitivity	Sens		-	20	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional, $I_P = 0\text{ A}$	-	$0.5 \times V_{CC}$	-	V
<b>TOTAL OUTPUT ERROR COMPONENTS [2] <math>E_{TOT} = E_{SENS} + 100 \times V_{OE} / (\text{Sens} \times I_P)</math></b>						
Total Output Error [3]	$E_{TOT}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$	-3.5	$\pm 1.8$	3.5	%
		$I_P = I_{PR(max)}$ , $T_A = 125^\circ\text{C}$	-3	$\pm 1.4$	3	%
		$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 2.9$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$	-6	$\pm 4$	6	%
Sensitivity Error	$E_{SENS}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = 125^\circ\text{C}$	-2.5	$\pm 1.6$	2.5	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$	-4.5	$\pm 3.1$	4.5	%
Voltage Offset Error	$V_{OE}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-30	$\pm 17$	30	mV
		$I_P = 0\text{ A}$ , $T_A = 125^\circ\text{C}$	-25	$\pm 7$	25	mV
		$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 28$	-	mV
		$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$	-70	$\pm 31$	70	mV
Linearity Error [4]	$E_{LIN}$	$T_A = 25^\circ\text{C}$ , up to full-scale $I_P$	-1.7	$\pm 1.1$	1.7	%
		$T_A = 125^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 0.5$	-	%
		$T_A = -40^\circ\text{C}$ , up to full-scale $I_P$	-	$\pm 2.8$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS [5] (lifetime drift limits apply after solder reflow)</b>						
Total Output Error Including Lifetime Drift	$E_{TOT(DRIFT)}$	$I_P = I_{PR(max)}$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 4$	-	%
		$I_P = I_{PR(max)}$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 4$	-	%
Sensitivity Error Including Lifetime Drift	$E_{SENS(DRIFT)}$	$I_P = I_{PR(max)}/2$ , $T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 2.2$	-	%
		$I_P = I_{PR(max)}/2$ , $T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 3.3$	-	%
Offset Voltage Error Including Lifetime Drift	$V_{OE(DRIFT)}$	$T_A = 25^\circ\text{C}$ , $125^\circ\text{C}$	-	$\pm 44$	-	mV
		$T_A = -40^\circ\text{C}$ , $25^\circ\text{C}$	-	$\pm 44$	-	mV

[1] Typical values with  $\pm$  are mean  $\pm 3$  sigma values, except for lifetime drift, which are the average value including drift after AEC-Q100 qualification.

[2] A single part does not have both the maximum sensitivity error and the maximum offset voltage, because that would violate the maximum/minimum total output error specification. For total error, 3-sigma distribution values for offset and sensitivity may be combined by taking the square root of the sum of the squares. See characteristic performance data plots for temperature drift performance.

[3] Percentage of  $I_P$ , with  $I_P = I_{PR(MAX)}$ .

[4] The sensor continues to respond to current beyond the range of  $I_{PR}$  until the high or low output saturation voltage. However, the nonlinearity in this region may be worse than the nominal operating range.

[5] Lifetime drift characteristics are based on AEC-Q100 qualification results. Typical values are mean  $\pm 3$  sigma of worst-case stress testing. Drift is a function of customer application conditions. Solder reflow induces stress on the ACS732 and ACS733; lifetime drift limits apply after solder reflow.

## CMTI APPLICATION INFORMATION

The ACS732/ACS733 was injected with 1000 V across the insulation barrier. The typical slew rate for which the devices settled within <100 mV within 0.3  $\mu$ s is noted in the Performance Characteristics table.

A pulse of 1000 V with a slew rate of 140 V/ns was injected into the device in order to determine the settling time to  $V_{OUT}$  within 100 mV. Figure 3 and Figure 4 below show the voltage disturbances for both positive and negative 1000 V injected pulses. Devices used typical application conditions and nominal  $V_{DD}$  values.

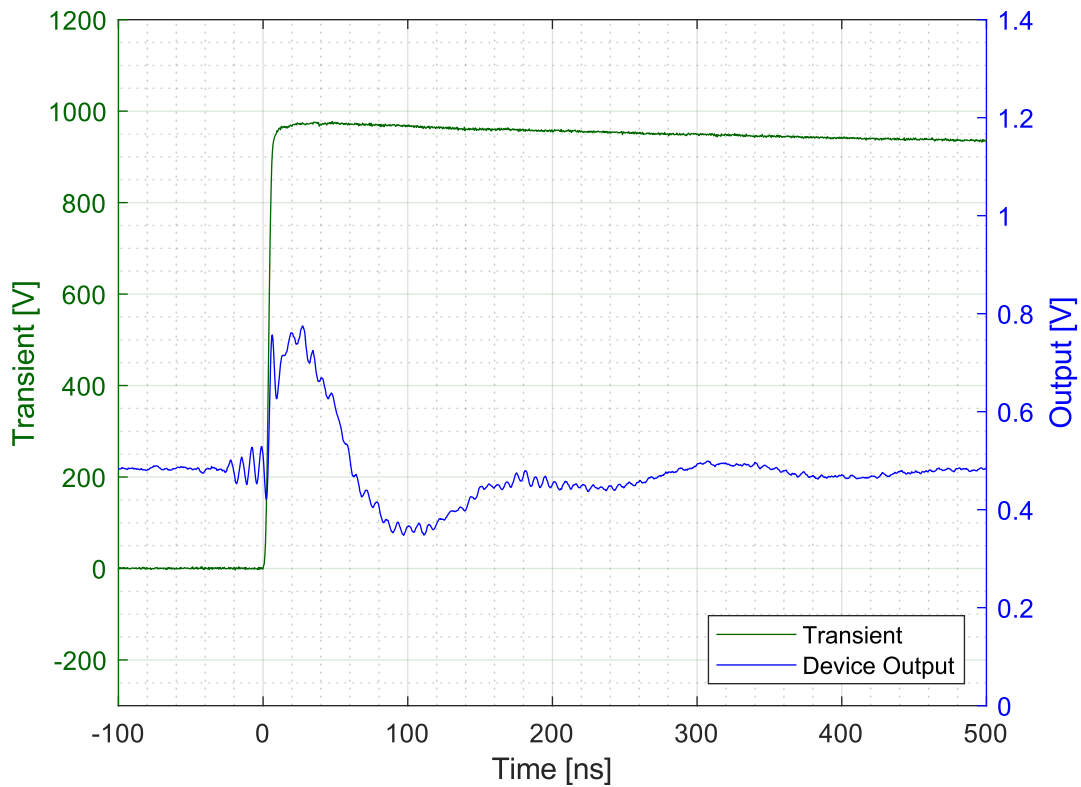


Figure 3: ACS732/ACS733 output response to positive 1000 V transient

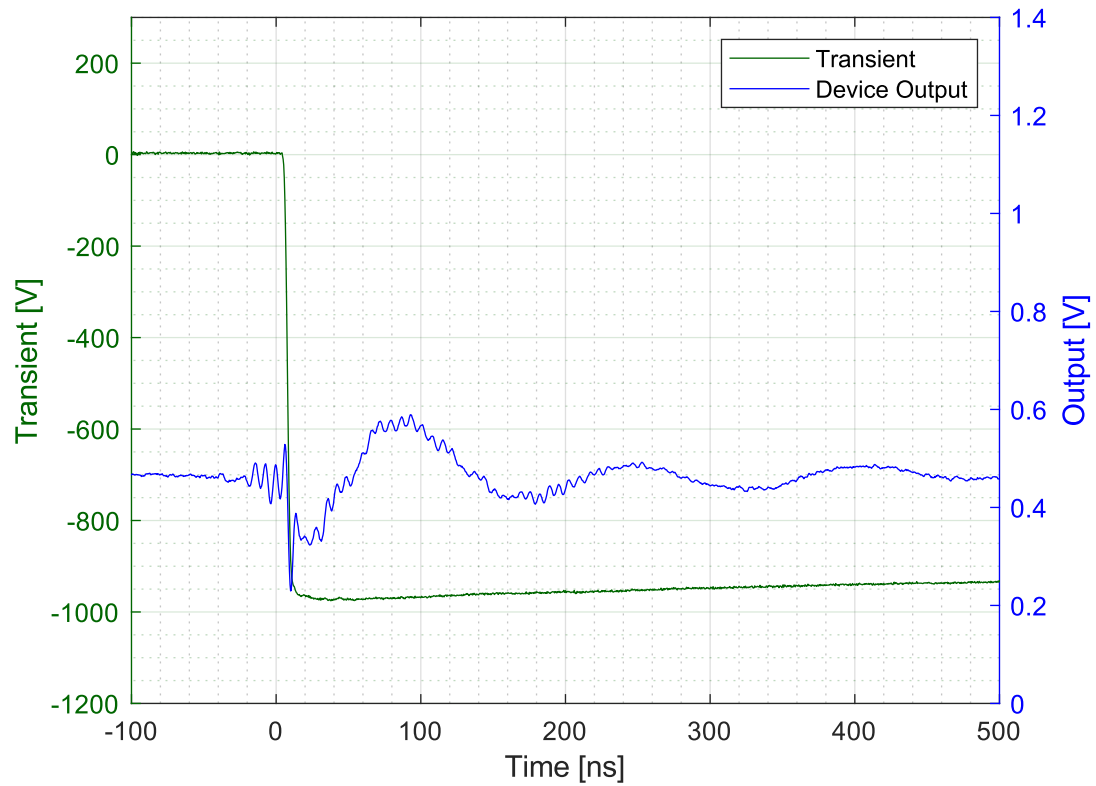
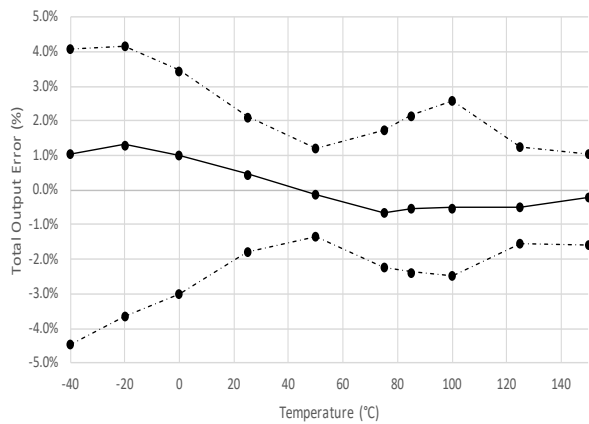
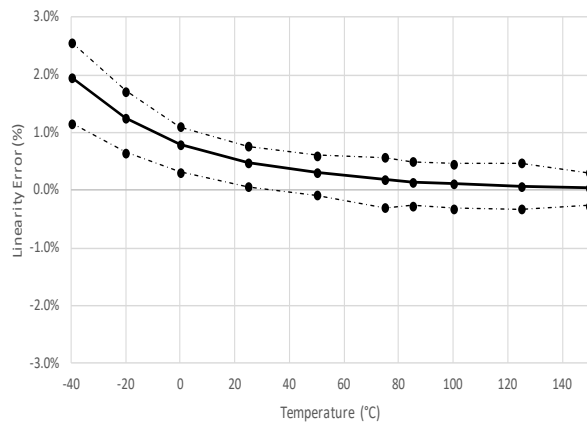
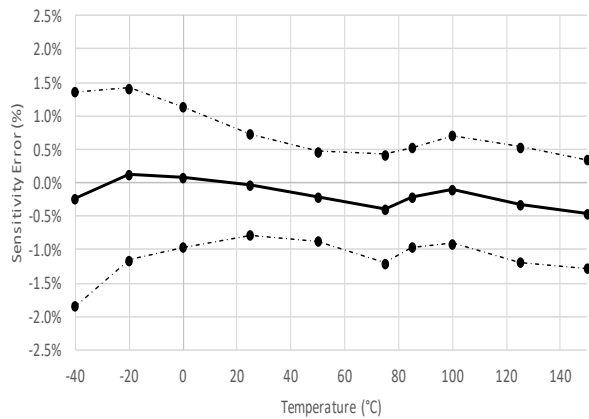
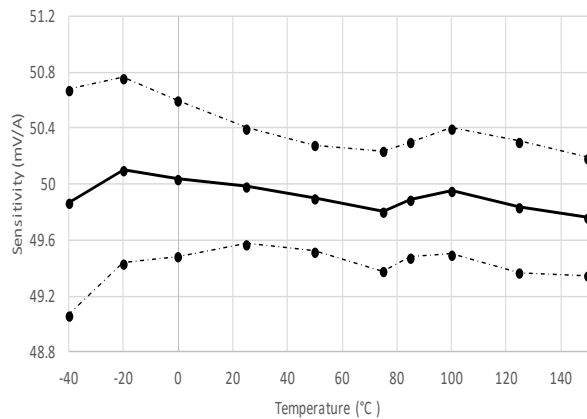
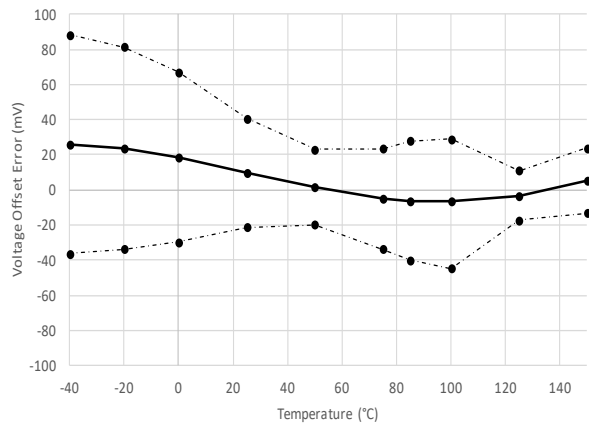
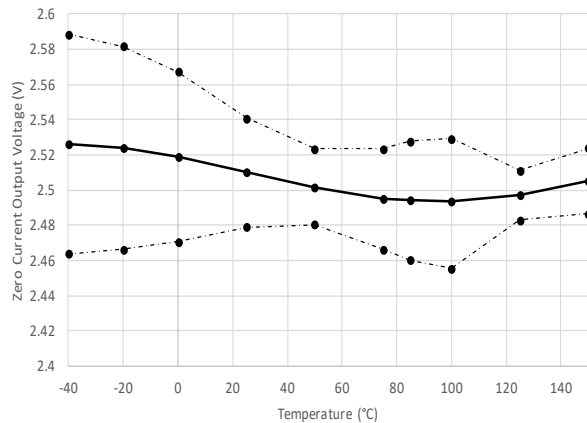


Figure 4: ACS732/ACS733 output response to negative 1000 V transient

## CHARACTERISTIC PERFORMANCE ACS732KLATR-40AB-T (5 V)

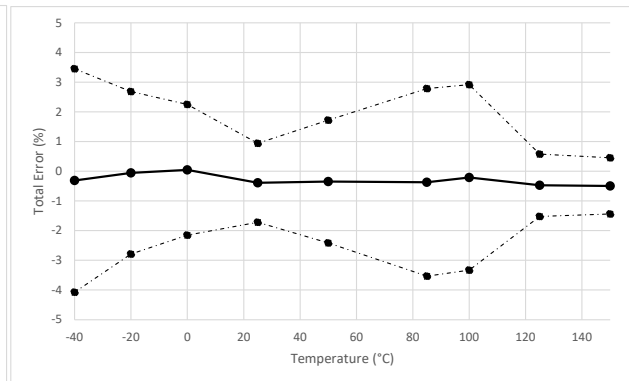
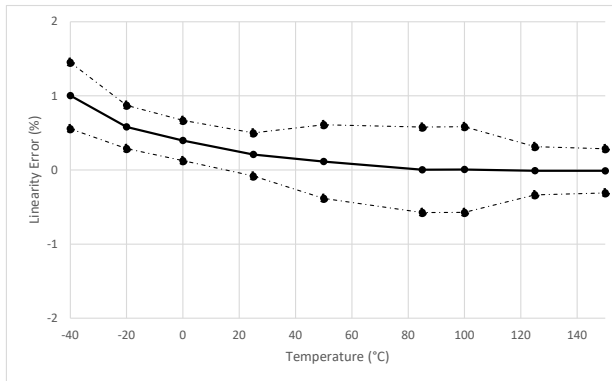
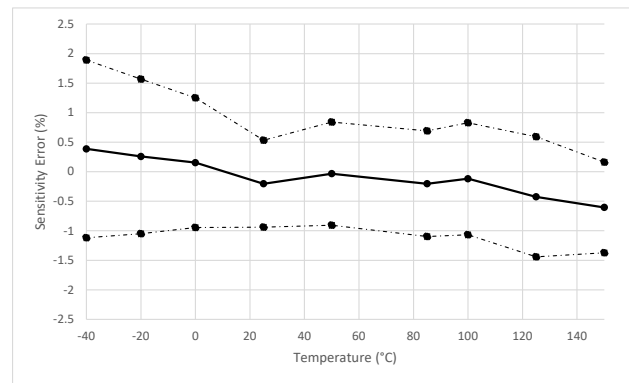
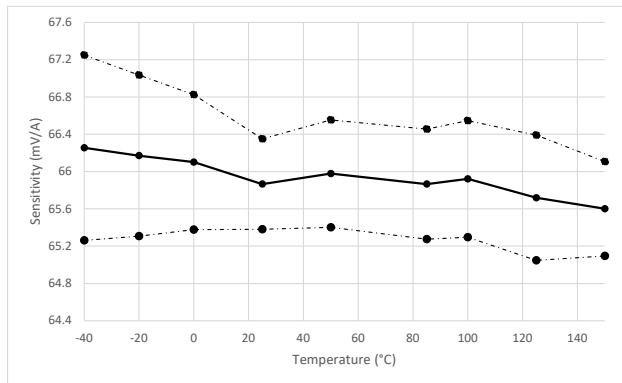
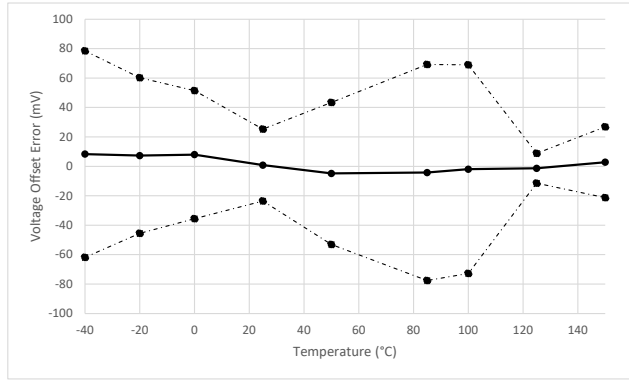
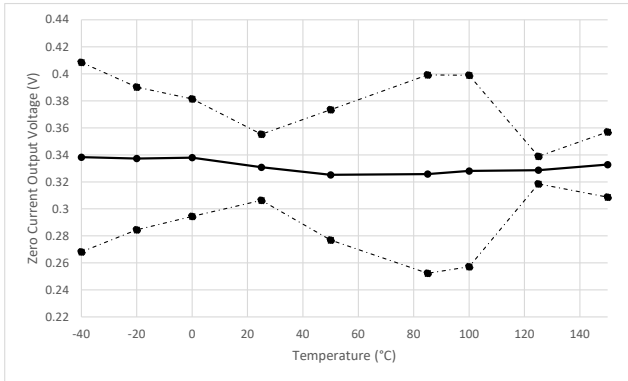
for ~150 parts



— μ    - - - μ ± 3σ

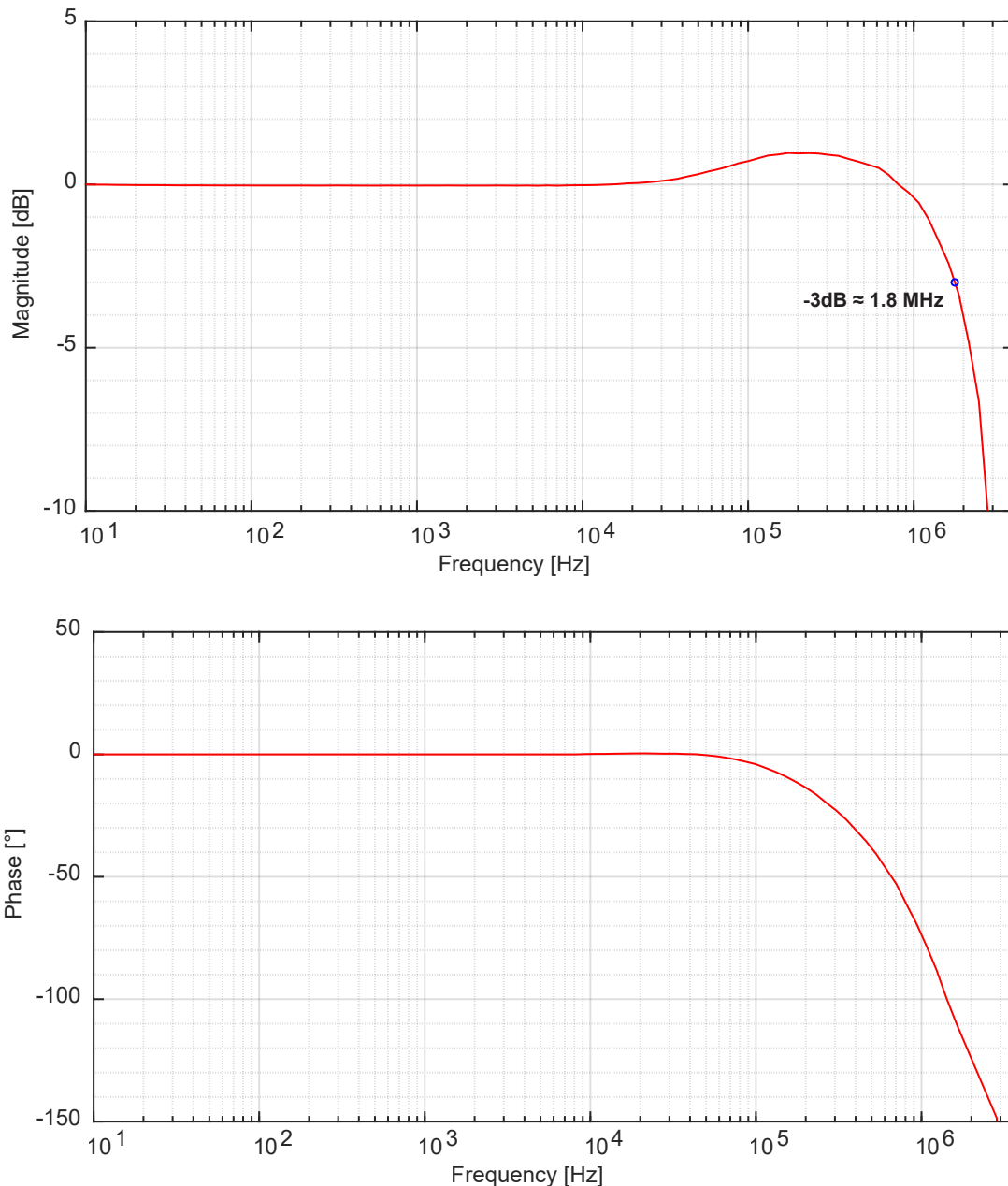
## CHARACTERISTIC PERFORMANCE ACS733KLA-40AU-T (3.3 V)

for ~50 parts



— μ    - - - μ ± 3σ

### CHARACTERISTIC PERFORMANCE ACS732 AND ACS733 TYPICAL FREQUENCY RESPONSE



For information regarding bandwidth characterization methods used for the ACS732 and ACS733, see the application note “An Effective Method for Characterizing System Bandwidth in Complex Current Sensor Applications” (<https://www.allegromicro.com/en/insights-and-innovations/technical-documents/hall-effect-sensor-ic-publications/an-effective-method-for-characterizing-system-bandwidth-an296169>) on the Allegro website.

## RESPONSE CHARACTERISTICS DEFINITIONS AND PERFORMANCE DATA

### Response Time ( $t_{\text{RESPONSE}}$ )

The time interval between a) when the sensed input current reaches 90% of its final value, and b) when the sensor output reaches 90% of its full-scale value.

### Propagation Delay ( $t_{\text{pd}}$ )

The time interval between a) when the sensed input current reaches 20% of its full-scale value, and b) when the sensor output reaches 20% of its full-scale value.

### Rise Time ( $t_r$ )

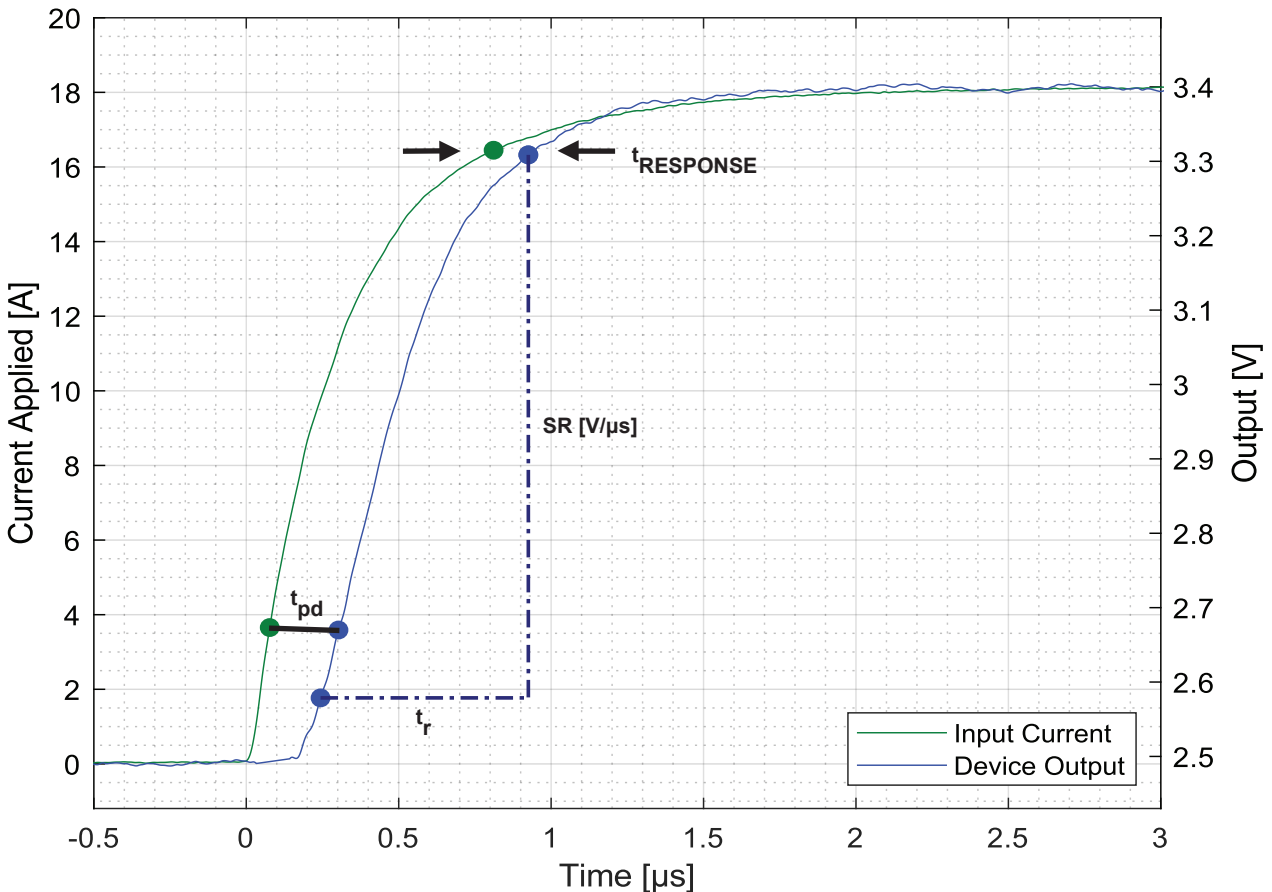
The time interval between a) when the sensor reaches 10% of its full-scale value, and b) when it reaches 90% of its full-scale value.

### Output Slew Rate (SR)

The rate of change [ $\text{V}/\mu\text{s}$ ] in the output voltage between a) when the sensor reaches 10% of its full-scale value, and b) when it reaches 90% of its full-scale value.

### Response Time, Propagation Delay, Rise Time, and Output Slew Rate (ACS732, 5 V)

Applied current step with 10% to 90% rise time =  $1 \mu\text{s}$   
 Test conditions:  $T_A = 25^\circ\text{C}$ ,  $C_{\text{BYPASS}} = 0.1 \mu\text{F}$ ,  $C_L = 0 \text{ F}$



## OVERCURRENT FAULT

### Overcurrent Fault

The ACS732 and ACS733 have fast and accurate overcurrent fault detection circuitry. The overcurrent fault threshold ( $I_{FAULT}$ ) is user-configurable via an external resistor divider and supports a range of 50% to 200% of the full-scale primary input ( $I_{PR(MAX)}$ ). Fault response and the overcurrent fault thresholds are described in the following sections.

### Fault Response

The high bandwidth of the ACS732 and ACS733 devices allow for extremely fast and accurate overcurrent fault detection. An overcurrent event occurs when the magnitude of the input current ( $I_p$ ) exceeds the user-set threshold ( $I_{FAULT}$ ). Fault response time ( $t_{RESPONSE(F)}$ ) is defined from the time  $I_p$  goes above  $I_{FAULT}$  to the time the  $\overline{FAULT}$  pin goes below  $V_{FAULT}$ . Overcurrent fault response is illustrated in Figure 5. When  $I_p$  goes below  $I_{FAULT} - I_{HYST}$ , the  $\overline{FAULT}$  pin is released. The rise time of  $V_{FAULT}$  depends on the value of the resistor  $R_{F(PULLUP)}$  and the capacitance on the pin.

### Setting the Overcurrent Fault Threshold

The overcurrent fault threshold ( $I_{FAULT}$ ) is set via a resistor divider from  $V_{CC}$  to ground on the VOC pin. The voltage on the VOC pin,  $V_{VOC}$ , may range from  $0.1 \times V_{CC}$  to  $0.4 \times V_{CC}$ .  $I_{FAULT}$  may be set anywhere from 50% to 200%  $I_{PR(MAX)}$ .

Overcurrent fault threshold versus  $V_{VOC}$  is shown in Figure 6.

The equation for calculating the trip current is shown below. For bidirectional devices, the fault trips for both positive and negative currents.

$$I_{FAULT} = I_{PR(MAX)} \left\{ 5 \times \frac{V_{VOC}}{V_{CC}} \right\}$$

This may be rearranged to solve for the appropriate  $V_{VOC}$  value based on a desired overcurrent fault threshold, shown by the equation:

$$V_{VOC} = \frac{V_{CC}}{5} \times \frac{I_{FAULT}}{I_{PR(MAX)}}$$

By setting  $V_{VOC}$  with a resistor divider from  $V_{CC}$ , the ratio of  $V_{VOC}/V_{CC}$  remains constant with changes to  $V_{CC}$ . In this regard, the fault trip point remains constant even as the supply voltage varies.

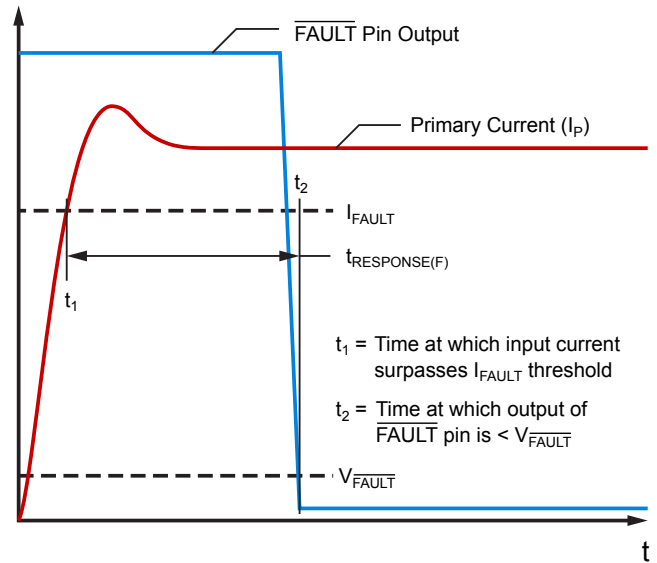


Figure 5: Overcurrent Fault Response

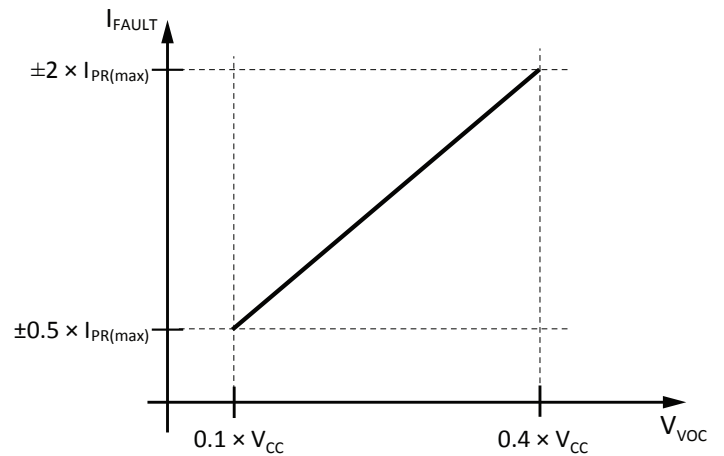


Figure 6: Fault Threshold vs.  $V_{VOC}$

It is best practice to use resistor values  $< 10 \text{ k}\Omega$  for setting  $V_{VOC}$ . With larger resistor values, the leakage current on VOC may result in errors in the trip point.



## DEFINITIONS OF ACCURACY CHARACTERISTICS

**Sensitivity (Sens).** The change in sensor IC output in response to a 1 A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) (1 G = 0.1 mT) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the full-scale current of the device.

**Nonlinearity ( $E_{LIN}$ ).** The nonlinearity is a measure of how linear the output of the sensor IC is over the full current measurement range. The nonlinearity is calculated as:

$$E_{LIN} = \left\{ 1 - \left[ \frac{V_{IOUT}(I_{PR(max)}) - V_{IOUT(Q)}}{2 \times V_{IOUT}(I_{PR(max)/2}) - V_{IOUT(Q)}} \right] \right\}$$

where  $V_{IOUT}(I_{PR(max)})$  is the output of the sensor IC with the maximum measurement current flowing through it and  $V_{IOUT}(I_{PR(max)/2})$  is the output of the sensor IC with half of the maximum measurement current flowing through it.

**Zero Current Output Voltage ( $V_{IOUT(Q)}$ ).** The output of the sensor when the primary current is zero. For a unipolar supply voltage, it nominally remains at  $0.5 \times V_{CC}$  for a bidirectional device and  $0.1 \times V_{CC}$  for a unidirectional device. For example, in the case of a bidirectional output device,  $V_{CC} = 3.3$  V translates into  $V_{IOUT(Q)} = 1.65$  V. Variation in  $V_{IOUT(Q)}$  can be attributed to the resolution of the Allegro linear IC quiescent voltage trim and thermal drift.

**Voltage Offset Error ( $V_{OE}$ ).** The deviation of the device output from its ideal quiescent value of  $0.5 \times V_{CC}$  (bidirectional) or  $0.1 \times V_{CC}$  (unidirectional) due to nonmagnetic causes. To convert this voltage to amperes, divide by the device sensitivity, Sens.

**Total Output Error ( $E_{TOT}$ ).** The difference between the current measurement from the sensor IC and the actual current ( $I_P$ ), relative to the actual current. This is equivalent to the difference between the ideal output voltage and the actual output voltage, divided by the ideal sensitivity, relative to the current flowing through the primary conduction path:

$$E_{TOT}(I_P) = \frac{V_{IOUT_{ideal}}(I_P) - V_{IOUT}(I_P)}{Sens_{ideal}(I_P) \times I_P} \times 100 (\%)$$

The Total Output Error incorporates all sources of error and is a function of  $I_P$ . At relatively high currents,  $E_{TOT}$  will be mostly due to sensitivity error, and at relatively low currents,  $E_{TOT}$  will be mostly due to Voltage Offset Error ( $V_{OE}$ ). As  $I_P$  approaches zero,  $E_{TOT}$  approaches infinity due to the offset voltage. This is illustrated in Figure 7 and Figure 8. Figure 7 shows a distribution of output voltages versus  $I_P$  at 25°C and across temperature. Figure 8 shows the corresponding  $E_{TOT}$  versus  $I_P$ .

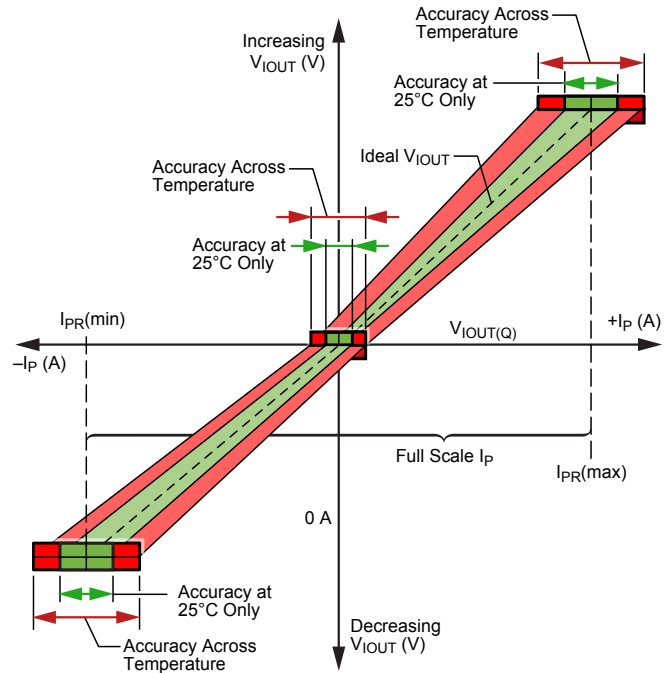


Figure 7: Output Voltage versus Sensed Current

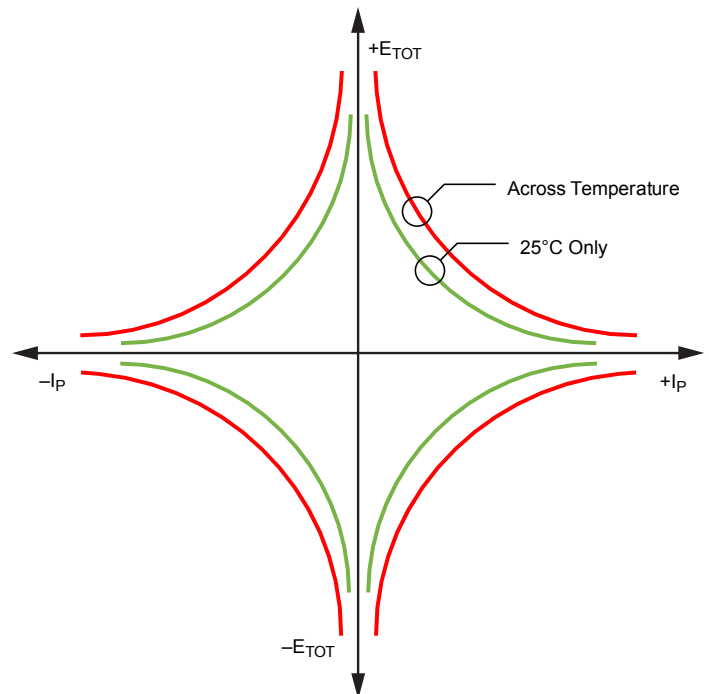


Figure 8: Total Output Error versus Sensed Current

## APPLICATION INFORMATION

### Ratiometry

The ACS732 and ACS733 are both ratiometric sensors. This means that, for a given change in supply voltage, the device zero-current output voltage and sensitivity scale proportionally.

### Sensitivity Ratiometry

Ideally, a 5% increase in  $V_{CC}$  results in a 5% increase in sensitivity. However, the ratiometric response of any sensor is not ideal. Ratiometric sensitivity error,  $E_{RAT(SENS)}$ , is specified by the equation:

$$E_{RAT(SENS)} = 100\% \times \left[ 1 - \left( \frac{Sensitivity_{V_{CC}}}{Sensitivity_{V_{CC(N)}}} \times \frac{V_{CC(N)}}{V_{CC}} \right) \right]$$

where  $V_{CC(N)}$  is equal to the nominal  $V_{CC}$  (3.3 V, or 5 V) and  $Sensitivity_{V_{CC(N)}}$  is the measured sensitivity at nominal  $V_{CC}$  for a particular device. The symbol  $V_{CC}$  is the measured  $V_{CC}$  value in application and  $Sensitivity_{V_{CC}}$  is the measured sensitivity at that  $V_{CC}$  level for a particular device.

### Zero-Current Offset Ratiometry

Ratiometric error for zero-current offset may be calculated using the following equation:

$$E_{RAT(Q)} = V_{IOUT(Q)V_{CC}} - V_{IOUT(Q)V_{CC(N)}} \times \frac{V_{CC}}{V_{CC(N)}}$$

where  $V_{CC(N)}$  is equal to the nominal  $V_{CC}$  (3.3 V, or 5 V) and  $V_{IOUT(Q)V_{CC(N)}}$  is the measured zero-current offset voltage at nominal  $V_{CC}$  for a particular device. The symbol  $V_{CC}$  is the measured  $V_{CC}$  value in application and  $V_{IOUT(Q)V_{CC}}$  is the measured zero-current offset voltage for a particular device.

### Estimating Total Error vs. Sensed Current

The performance characteristics tables provide distribution ( $\pm 3$ -sigma) values for total error at  $I_{PR(MAX)}$ ; however, one may be interested in the expected error at a particular current. This error may be estimated using the distribution data for the components of total error, sensitivity error, and offset voltage. The  $\pm 3$ -sigma value for total error ( $E_{TOT}$ ) as a function of the sensed current is estimated as:

$$E_{TOT}(I_p) = \sqrt{E_{SENS}^2 + \left( \frac{100 \times V_{OE}}{Sens \times I_p} \right)^2}$$

where  $E_{SENS}$  and  $V_{OE}$  are the  $\pm 3$ -sigma values for those error terms.

If there is an average sensitivity error or average offset voltage, the average total error is estimated as:

$$E_{TOTAVG}(I_p) = E_{SENSAVG} + \frac{100 \times V_{OEAVG}}{Sens \times I_p}$$

### Layout Guidelines

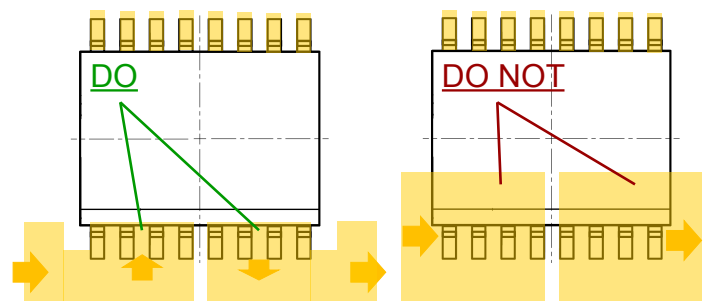
There are a few considerations during printed circuit board (PCB) layout that help to maintain high accuracy when using Allegro integrated current sensors. Common layout mistakes that should be avoided are:

- Extending current-carrying traces too far beneath the IC, or injecting current from the side of the IC
- Placing secondary-current phase traces too close to or below the IC

### Extending the Current Traces

The length of copper trace beneath the IC may impact the path of current flowing through the IP bus. This may cause variation in the coupling factor from the primary current loop of the package to the IC and may reduce the overall creepage distance in application.

It is best practice for the current to approach the IC parallel to the current-carrying pins and for the current-carrying trace to not creep toward the center of the package. Refer to Figure 9.



**Figure 9: Best Practice Layout Techniques for Current Traces**

If current must approach the package from the side, it is recommended to reduce the angle as much as possible. For more information about best current sensor layout practices, refer to the application note “[Techniques to Minimize Common-Mode Field Interference When Using Allegro Current Sensor ICs](#)” on the Allegro website.

## Thermal Rise vs. Primary Current

Self-heating due to the flow-off current should be considered during the design of any current sensing system. The sensor, PCB, and contacts to the PCB generate heat as current moves through the system.

The thermal response is highly dependent on PCB layout, copper thickness, cooling techniques, and the profile of the injected current. The current profile includes peak current, current “on-time,” and duty cycle. While the data presented in this section was collected with direct current (DC), these numbers may be used to approximate thermal response for both AC signals and current pulses.

The plot in Figure 10 shows the measured rise in steady-state die temperature of the ACS732/3 versus continuous current at an ambient temperature,  $T_A$ , of 25°C. The thermal offset curves may be directly applied to other values of  $T_A$ . Conversely, Figure 11 shows the maximum continuous current at a given  $T_A$ . Surges beyond the maximum current are allowed given the maximum junction temperature,  $T_{J(MAX)}$  (165°C), is not exceeded.

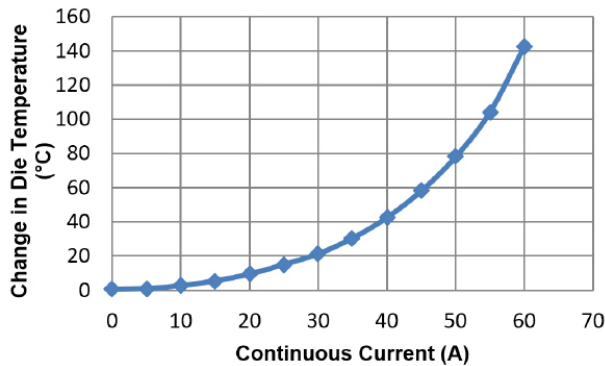


Figure 10: Self-Heating in the LA Package Due to Current Flow

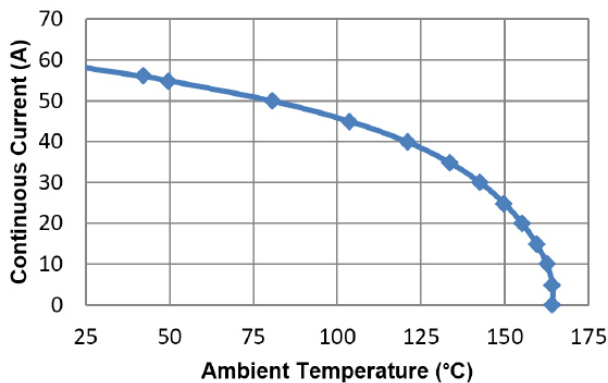


Figure 11: Maximum Continuous Current at a Given  $T_A$

The thermal capacity of the ACS732/3 should be verified by the end user in the application-specific conditions. The maximum junction temperature,  $T_{J(MAX)}$  (165°C), should not be exceeded. Further information on this application testing is available in the application note "DC and Transient Current Capability" on the Allegro website.

## ASEK73x Evaluation Board Layout

Thermal data was collected using the ASEK73x evaluation board (TED-0001795). This board includes 1500 mm<sup>2</sup> of 2 oz. (0.0694 mm) copper connected to pins 1 through 4 and pins 5 through 8, with thermal vias connecting the layers. Top and bottom layers of the PCB are shown in Figure 12.

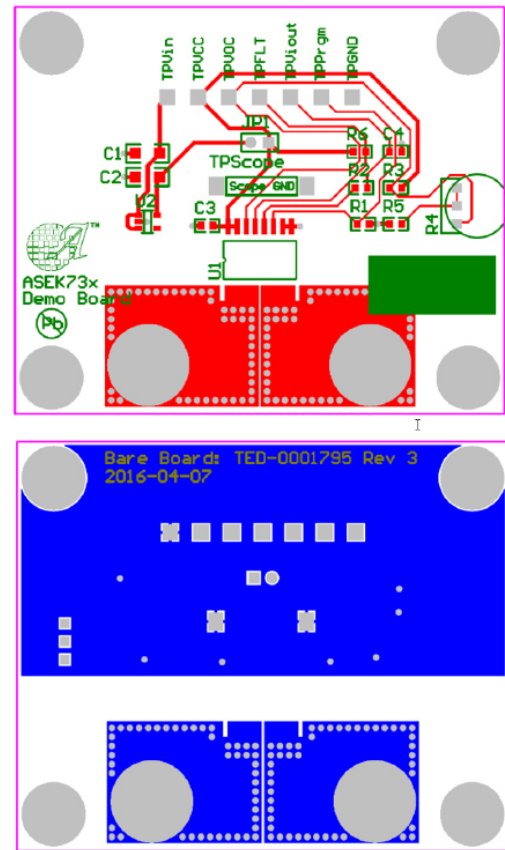


Figure 12: Top and Bottom Layers for ASEK73x Evaluation Board

Gerber files for the ASEK73x evaluation board are available for download from the Allegro website. See the technical documents section of the [ACS732 and ACS733 device webpage](#).

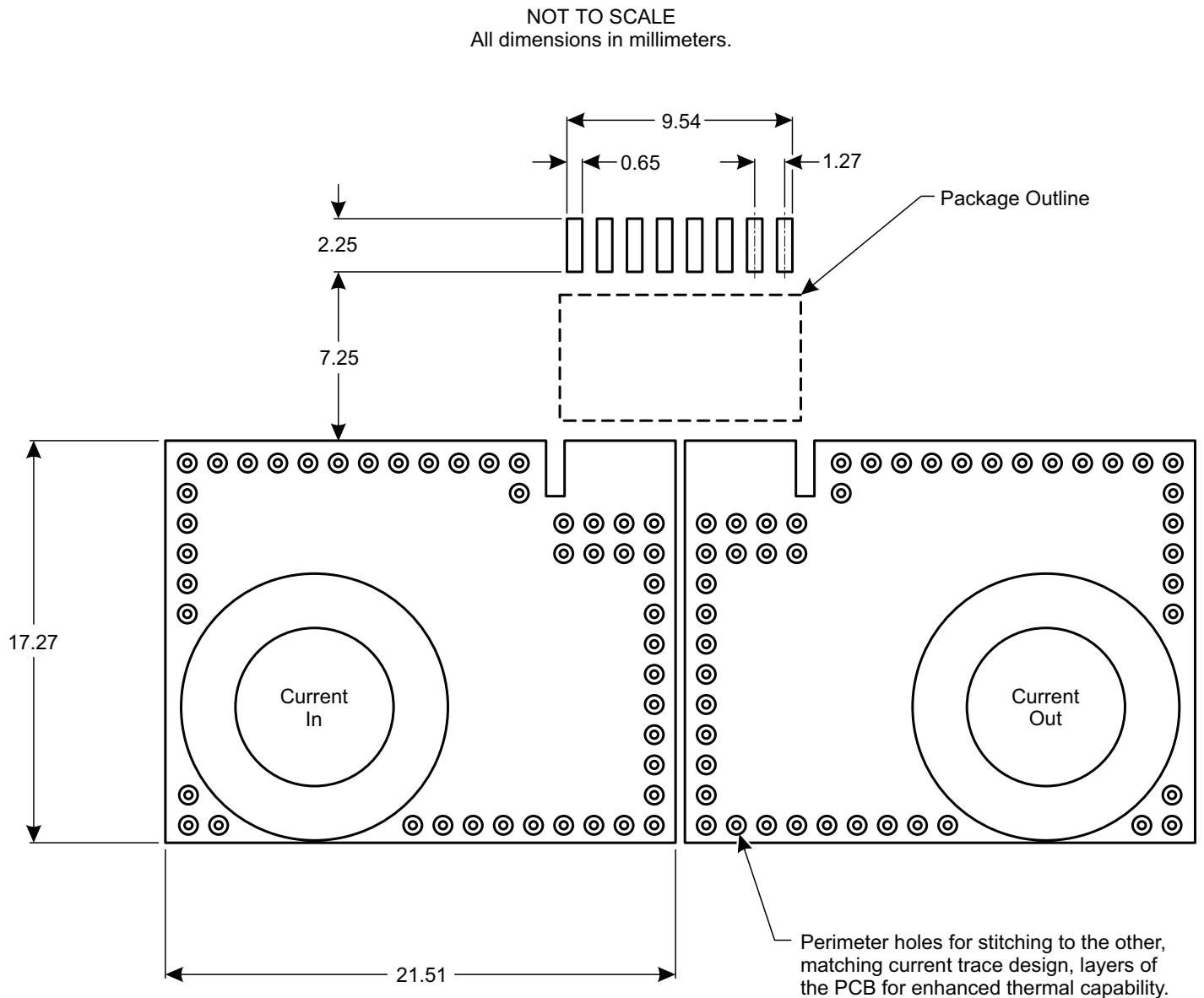


Figure 13: High-Isolation PCB Layout

## PACKAGE OUTLINE DRAWING

### For Reference Only – Not for Tooling Use

(Reference Allegro DWG-0000388, Rev. 1 and JEDEC MS-013AA)  
 NOT TO SCALE  
 Dimensions in millimeters  
 Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown

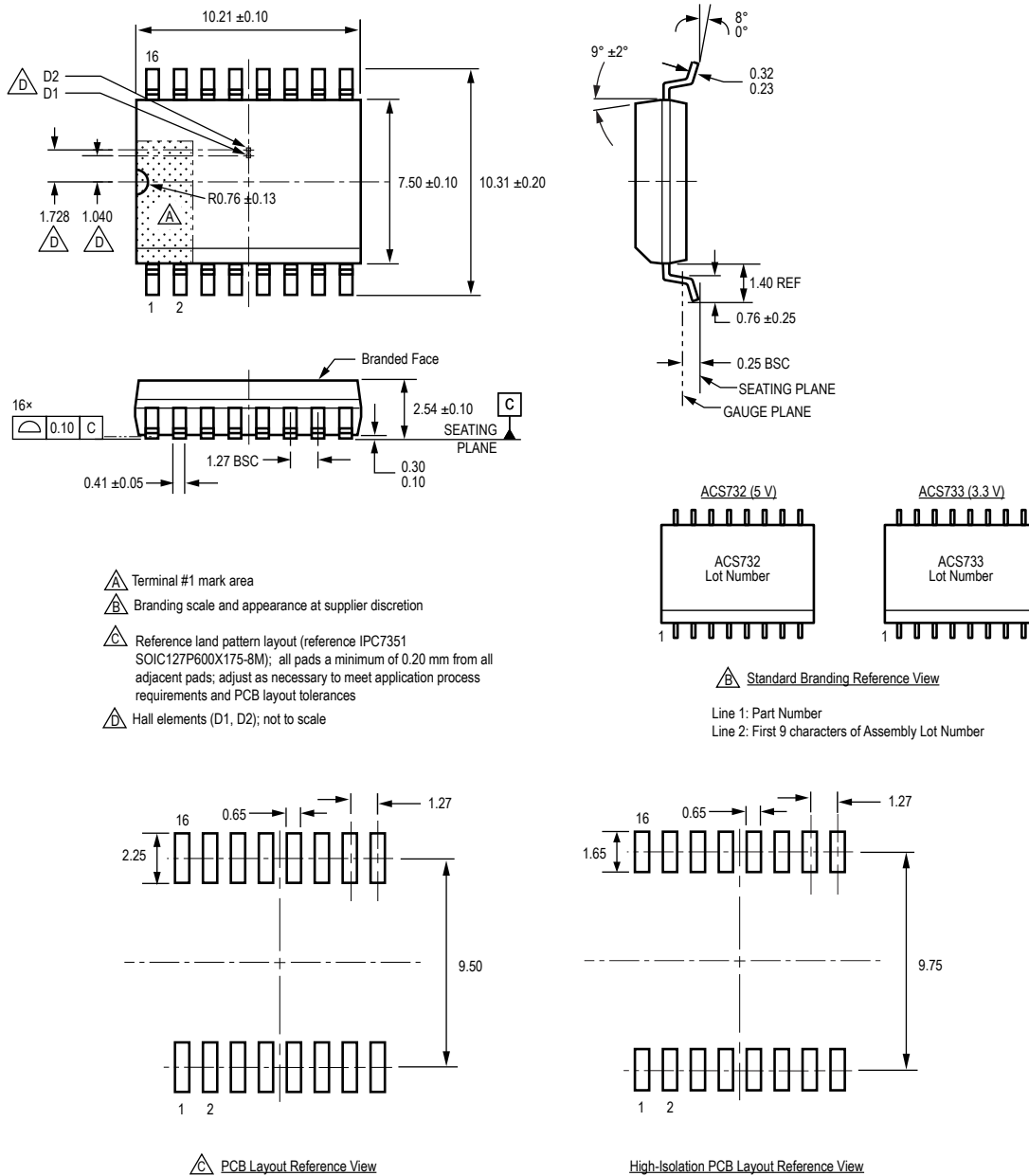


Figure 14: Package LA, 16-Pin SOICW

## Revision History

Number	Date	Description
–	September 20, 2017	Initial release
1	January 8, 2018	Updated Rise Time, Response Time, and Propagation Delay Time (page 5)
2	March 8, 2018	Added ACS732KLATR-20AB-T part option
3	June 20, 2018	Updated Working Voltage for Basic Isolation units (page 4); added Fault Response Time and Fault Release Time characteristics (page 6)
4	July 2, 2018	Added “Thermal Rise vs. Primary Current” and “ASEK73x Evaluation Board Layout” to the Applications Information section (page 28)
5	October 1, 2018	Added ACS732KLATR-75AB-T variant (pages 2 and 9) Updated Secondary Hall Coupling Factor value (page 5)
6	November 13, 2018	Added ACS732KLATR-65AU-T part option (page 2 and 10) Added ACS732KLATR-75AB-T characteristic performance plots (page 18)
7	November 16, 2018	Added ACS732KLATR-65AB-T part option (page 2, 10, and 19)
8	December 10, 2018	Updated UL certificate number
9	February 26, 2019	Added Dielectric Surge Strength Test Voltage to Isolation Characteristics table (page 3)
10	May 23, 2019	Updated Sensitivity Error (pages 10-11) and Total Output Error (page 11)
11	August 22, 2019	Added Maximum Continuous Current to Absolute Maximum Ratings table (page 3), ESD ratings table (page 3), and updated thermal data section (page 34)
12	September 10, 2019	Added Hall plate dimensions (page 36)
13	September 26, 2019	Added Hall element positions to package outline drawings (page 36)
14	January 17, 2020	Corrected Reverse $V_{OC}$ Voltage value (page 3); added Distance Through Insulation and Comparative Tracking Index to Isolation Characteristics table (page 3); updated Rise Time, Response Time, Propagation Delay, and Output Slew Rate test conditions, and added Output Slew Rate (page 6); removed Characteristic Performance plots (pages 17-24); updated Typical Frequency Response plots (page 17)
15	May 8, 2020	Removed Linearity Error from Common Electrical Characteristics table and added Linearity Error to Performance Characteristics Tables (pages 8-16); corrected Sensitivity Error test conditions (pages 8-16); corrected Lifetime Drift Characteristics numbers (pages 8-16); added ACS732KLATR-40AB-T multi-temperature characteristic performance plots (page 17); updated Typical Frequency Response plots (page 18); added Response Characteristics Definitions and Performance Data application page (page 19)
16	May 28, 2020	Updated Hall placement (p. 25)
17	June 9, 2020	Added ACS733KLATR-20AB-H part option (page 2, 14); updated Features and Benefits
18	June 29, 2020	Added minimum and maximum values to Linearity Error at $T_A = 25^\circ\text{C}$ (pages 8-17)
19	August 3, 2020	Corrected ACS732KLATR-40AB Total Output Error, Sensitivity Error, and Voltage Offset Error numbers; corrected ACS733KLATR-20AB and -20AB-H Total Output Error Including Lifetime Drift numbers
20	September 17, 2021	Updated Output Saturation Voltage $V_{SAT(LOW)}$ maximum values (page 6); added ACS733KLA-40AU-T characteristic performance plots (page 19)
21	October 7, 2022	Updated package drawing (page 27)
22	September 21, 2023	Updated Isolation Characteristics (page 3)
23	January 19, 2024	Updated Clearance and Creepage values (page 3) and Thermal Characteristics (page 4)
24	February 7, 2024	Removed Maximum Continuous Current (page 3); updated ACS733KLA-40AU-T performance characteristics (page 16); updated footnote [5] of device performance tables (page 8–17); minor editorial updates (all pages)
25	September 26, 2024	Updated Isolation Characteristics (page 1 and 3); added Common Mode Transient Immunity application information (page 6, 18-19); updated lifetime drift characteristics header in Performance Characteristics tables (pages 8-17)

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