

#### FEATURES AND BENEFITS

- Integrated capacitor reduces requirements for external EMI-protection components
- Two-wire, pulse-width output protocol
- Highly configurable output protocol options
- · Speed and direction information of target
- Vibration tolerance
  - ☐ Small signal lockout for small amplitude vibration
  - ☐ Proprietary vibration detection algorithms for large amplitude vibration
- · Air-gap-independent switch points
- · Undervoltage lockout
- True zero-speed operation
- · Wide operating voltage range
- · Single chip sensing IC for high reliability
- Robust test-coverage capability with Scan Path and IDDQ measurement
- · Integrated back-biasing magnet

## PACKAGE: 3-pin SIP (suffix SN)



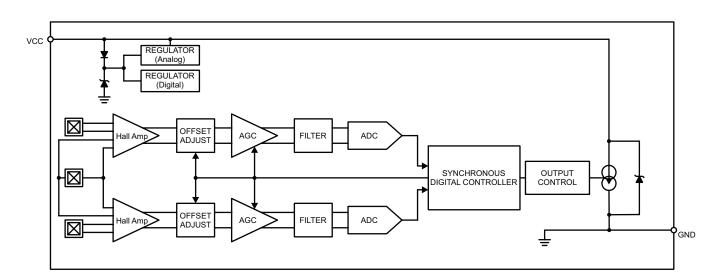
#### DESCRIPTION

The ATS699LSN is an optimized Hall-effect integrated circuit (IC) and rare-earth pellet combination that provides a user-friendly solution for direction detection and true zero-speed, digital gear-tooth sensing. The small package can be easily assembled and used in conjunction with a wide variety of gear-tooth-sensing applications.

The IC employs patented algorithms for the special operational requirements of automotive transmission applications. The speed and direction of the target are communicated through a variable pulse-width output protocol. The ATS699 is particularly adept at handling vibration without sacrificing maximum air gap capability or creating any erroneous "direction" pulses. Even higher angular vibration caused by engine cranking is completely rejected by the device. The advanced vibration detection algorithm will systematically calibrate the sensor IC on the initial teeth of true target rotation and not on vibration, always guaranteeing an accurate signal in running mode.

Advanced signal processing and innovative algorithms make the ATS699 an ideal solution for a wide range of speed and direction-sensing needs.

This device is available in a lead (Pb) free 3-pin SIP package with tin-plated leadframe.



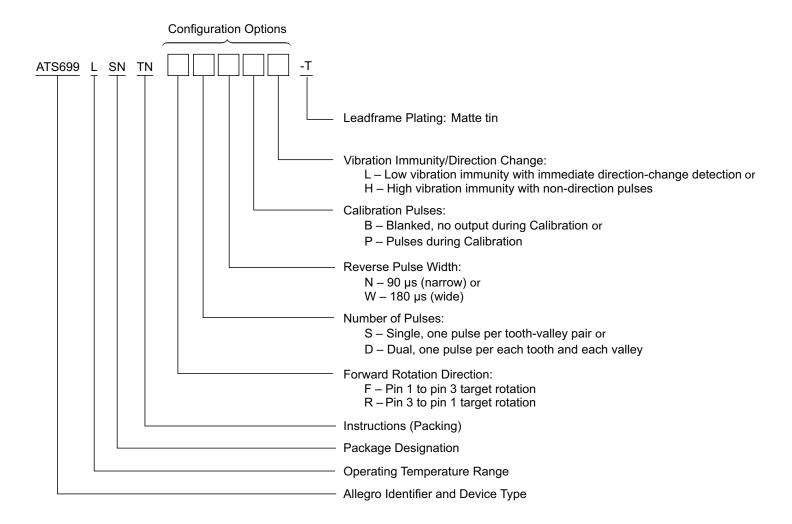
**Functional Block Diagram** 

## Two-Wire, Differential, Vibration-Resistant Sensor with Speed and Direction Output

### **SELECTION GUIDE**

Part Number	Packing
ATS699LSNTN-FSWPH-T	Tape and reel, 13-in. reel, 800 pieces per reel
ATS699LSNTN-RSWPH-T	Tape and reel, 13-in. reel, 800 pieces per reel

<sup>\*</sup> Not all combinations are available. Contact Allegro sales for availability and pricing of custom programming options.



For example: ATS699LSNTN-RSNPL-T

Where a configuration character is unspecified, "x" will be used. For example, -xSNPL applies to both Rotation Direction configuration variants.

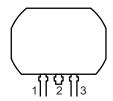


## **SPECIFICATIONS**

### **Absolute Maximum Ratings**

Characteristic	Symbol	Notes	Rating	Units
Supply Voltage	V <sub>CC</sub>	Refer to Power Derating section	28	V
Reverse Supply Voltage	V <sub>RCC</sub>		-18	V
Operating Ambient Temperature	T <sub>A</sub>	Range L	-40 to 150	°C
Maximum Junction Temperature	T <sub>J(max)</sub>		165	°C
Storage Temperature	T <sub>stg</sub>		-60 to 170	°C

## **Pinout Diagram and Terminal List**



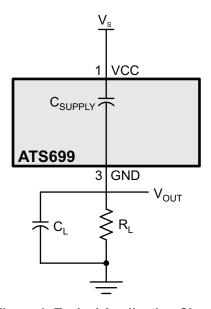
Package SN, 3-Pin SIP Pinout Diagram

#### **Terminal List Table**

Number	Name	Function			
1	VCC	Supply voltage			
2	VCC	Supply voltage			
3	GND	Ground			

### **Internal Discrete Capacitor Ratings**

Characteristic	Symbol	Notes	Rating	Units
Nominal Capacitance	C <sub>SUPPLY</sub>	Connected between VCC and GND	10000	pF



**Figure 1: Typical Application Circuit** 



# Two-Wire, Differential, Vibration-Resistant Sensor with Speed and Direction Output

## OPERATING CHARACTERISTICS: valid throughout full operating and temperature ranges; using Reference Target 60-0, unless otherwise noted.

Characteristics	Symbol	Test Conditions	Min.	Typ. [1]	Max.	Unit
ELECTRICAL CHARACTERISTICS						
Supply Voltage [2]	V <sub>CC</sub>	Operating, T <sub>J</sub> < T <sub>J</sub> (max)	4	_	24	V
Undervoltage Lockout	V <sub>CC(UV)</sub>	$V_{CC} 0 \rightarrow 5 V \text{ or } 5 \rightarrow 0 V$	_	3.6	3.95	V
Reverse Supply Current	I <sub>RCC</sub>	V <sub>CC</sub> = V <sub>RCC (MAX)</sub>	_	_	-10	mA <sup>[3]</sup>
Supply Zener Clamp Voltage	V <sub>ZSUPPLY</sub>	I <sub>CC</sub> = I <sub>CC(HIGH)</sub> + 3 mA, T <sub>A</sub> = 25°C	28	-	-	V
	I <sub>CC(Low)</sub>	Low-current state (Running mode)	5	-	8	mA
Supply Current	I <sub>CC(High)</sub>	High-current state (Running mode)	12	_	16	mA
	I <sub>CC(SU)(Low)</sub>	Startup current level and Power-On mode	5	_	8.5	mA
Supply Current Ratio	I <sub>CC(High)</sub> /I <sub>CC(Low)</sub>	Measured as a ratio of high current to low current	1.9	_	-	_
OUTPUT STAGE						
Output Rise Time	t <sub>r</sub>	$\Delta$ I/ $\Delta$ t from 10% to 90% I <sub>CC</sub> level; corresponds to measured output slew rate with C <sub>SUPPLY</sub>	0	2	4	μs
Output Fall Time	t <sub>r</sub>	$\Delta$ I/ $\Delta$ t from 90% to 10% I <sub>CC</sub> level; corresponds to measured output slew rate with C <sub>SUPPLY</sub>	0	2	4	μs
OUTPUT PULSE CHARACTERISTI	CS [4]					
Pulse Width, Forward Rotation	t <sub>w(FWD)</sub>		38	45	52	μs
Pulse Width, Reverse Rotation		-xxNxx variant	76	90	104	μs
Pulse Width, Reverse Rotation	t <sub>w(REV)</sub>	-xxWxx variant	153	180	207	μs
Pulse Width, Non-Direction	t <sub>w(ND)</sub>	-xxNPx and -xxNxH variants	153	180	207	μs
Pulse Width, Non-Direction		-xxWPx and -xxWxH variants	306	360	414	μs
PERFORMANCE CHARACTERISTI	cs					
Operate Point	B <sub>OP</sub>	% of peak-to-peak V <sub>PROC</sub>	_	69	_	%
Release Point	B <sub>RP</sub>	% of peak-to-peak V <sub>PROC</sub>	_	31	_	%
Operating Frequency, Forward	f	-xSxxx variant	0	_	12	kHz
Rotation	f <sub>FWD</sub>	-xDxxx variant	0	_	6	kHz
		-xSNxx variant	0	_	7	kHz
Operating Frequency, Reverse	f	-xDNxx variant	0	_	3.5	kHz
Rotation	f <sub>REV</sub>	-xSWxx variant	0	_	4	kHz
		-xDWxx variant	0	_	2	kHz
		-xSNxx variant	0	_	4	kHz
Operating Frequency, Non-Direction	f	-xDNxx variant	0	_	2	kHz
Pulses [5]	f <sub>ND</sub>	-xSWxx variant	0	_	2.2	kHz
		-xDWxx variant	0	_	1.1	kHz

Continued on the next page...

<sup>&</sup>lt;sup>5</sup> Maximum Operating Frequency is determined by satisfactory separation of output pulses:  $I_{CC(LOW)}$  of  $tw_{(FWD)(MIN)}$ . If the customer can resolve shorter low-state durations, maximum  $f_{REV}$  and  $f_{ND}$  may be increased.



 $<sup>^{1}</sup>$  Typical values are at  $T_A = 25$ °C and  $V_{CC} = 12$  V. Performance may vary for individual units, within the specified maximum and minimum limits.

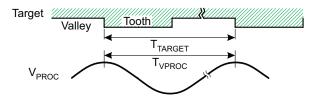
<sup>&</sup>lt;sup>2</sup> Maximum voltage must be adjusted for power dissipation and junction temperature; see Power Derating section.

<sup>&</sup>lt;sup>3</sup> Negative current is defined as conventional current coming out of (sourced from) the specified device terminal.

 $<sup>^4</sup>$  Load circuit is  $R_L$  = 100  $\Omega$  and  $C_L$  = 10 pF. Pulse duration measured at threshold of (  $(I_{CC(HIGH)}+I_{CC(LOW)})$  /2)

## OPERATING CHARACTERISTICS (continued): valid throughout full operating and temperature ranges; using Reference Target 60-0, unless otherwise noted.

Characteristics	Symbol	Test Co	Min.	Typ. [1]	Max.	Unit			
DAC CHARACTERISTICS	DAC CHARACTERISTICS								
Allowable User-Induced Offset		Magnitude valid for both channels	-60	-	60	G			
PERFORMANCE CHARACTERISTICS	PERFORMANCE CHARACTERISTICS								
Operational Magnetic Range	B <sub>IN</sub>	Peak to peak differentia magnetic channel.	30	-	1200	G			
Air Gap Range		Using Allegro 60-0 refe	0.5	_	2.5	mm			
Vilamatica Images units (Chambur)			-xxxxL variant	T <sub>TARGET</sub>	_	-	degrees		
Vibration Immunity (Startup)	Err <sub>VIB(SU)</sub>	See Figure 2	-xxxxH variant	T <sub>TARGET</sub>	_	-	degrees		
Vibration Immunity (Running Mode)	inning Mode) Err <sub>VIB</sub>	See Figure 2	-xxxxL variant	0.12 × T <sub>TARGET</sub>	_	_	degrees		
			-xxxxH variant	T <sub>TARGET</sub>	_	_	degrees		



 $V_{\tiny{PROC}}$  = the processed analog signal of the sinusoidal magnetic input (per channel)

 $T_{\text{TARGET}}$  = the period between successive similar (rising or falling) sensed magnetic edges

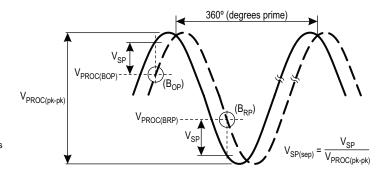


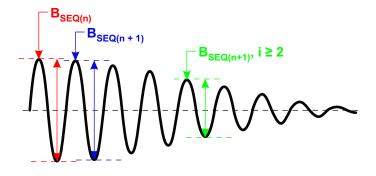
Figure 2: Definition of T<sub>TARGET</sub>

<sup>&</sup>lt;sup>1</sup> Typical values are at  $T_A = 25^{\circ}$ C and  $V_{CC} = 12$  V. Performance may vary for individual units, within the specified maximum and minimum limits.



## OPERATING CHARACTERISTICS (continued): valid throughout full operating and temperature ranges; using Reference Target 60-0, unless otherwise noted.

Characteristics	Symbol	Test Co	Min.	Typ. [1]	Max.	Unit		
INPUT MAGNETIC CHARACTERISTICS								
Allowable Differential Sequential Signal Variation 1	B <sub>SEQ(n+1)</sub> / B <sub>SEQ(n)</sub>	Signal cycle-to-cycle va	riation (see Figure 3)	0.6	_	_	_	
Allowable Differential Sequential Signal Variation 2	B <sub>SEQ(n+1)</sub> / B <sub>SEQ(n)</sub>	Overall signal variation	(see Figure 3)	0.4	_	-	-	
CALIBRATION								
		Amount of target rotation (constant direction) following	B <sub>IN</sub> > 60 G <sub>PP</sub> B <sub>IN</sub> ≤ 1200 G <sub>PP</sub>	_	2 × T <sub>TARGET</sub>	< 3 × T <sub>TARGET</sub>	degrees	
First Direction Output Pulse [2]	of either t <sub>w(FWD)</sub> or	power-on until lirst electrical output pulse of either $t_{\text{W(FWD)}}$ or $t_{\text{W(REV)}}$ . See Figure 2.	30 G <sub>PP</sub> ≤ B <sub>IN</sub> B <sub>IN</sub> ≤ 60 G <sub>PP</sub>	_	2.5 × T <sub>TARGET</sub>	< 4 × T <sub>TARGET</sub>	degrees	
First Direction-Pulse Output Following Direction Change	NCD	Amount of target rotation (constant direction) following event until first	-xxxxL variant	_	1	_	switch- point	
	NOD	electrical output pulse of either $t_{w(FWD)}$ or $t_{w(REV)}$ . $V_{SP(sep)} \ge 35$ . See Figure 2	-xxxxH variant	1 × T <sub>TARGET</sub>	2 × T <sub>TARGET</sub>	< 3 × T <sub>TARGET</sub>	degrees	
First Direction-Pulse Output Following Running Mode Vibration	rotati direc	Amount of target rotation (constant direction) following event until first	-xxxxL variant	_	_	1.25 × T <sub>TARGET</sub>	degrees	
		electrical output pulse of either $t_{w(FWD)}$ or $t_{w(REV)}$ . See Figure 2	-xxxxH variant	1 × T <sub>TARGET</sub>	2 × T <sub>TARGET</sub>	< 3 × T <sub>TARGET</sub>		



**Figure 3: Differential Signal Variation** 

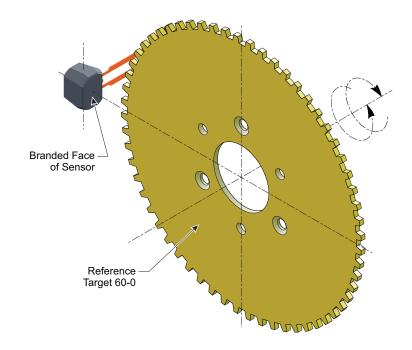
<sup>&</sup>lt;sup>2</sup> Power-up frequencies ≤ 200 Hz. Higher power-on frequencies may require more input magnetic cycles until output edges are achieved.



<sup>&</sup>lt;sup>1</sup> Typical values are at T<sub>A</sub> = 25°C and V<sub>CC</sub> = 12 V. Performance may vary for individual units, within the specified maximum and minimum limits.

## Reference Target 60-0 (60 Tooth Target)

Characteristics	Symbol	Test Conditions	Тур.	Units	Symbol Key
Outside Diameter	D <sub>o</sub>	Outside diameter of target	120	mm	$^{\varnothing D_0} \setminus {}^{h_t} \setminus F \longrightarrow F$
Face Width	F	Breadth of tooth, with respect to branded face	6	mm	
Circular Tooth Length	t	Length of tooth, with respect to branded face	3	degrees	Branded Face of Package
Circular Valley Width	t <sub>v</sub>	Length of valley, with respect to branded face	3	degrees	
Tooth Whole Depth	h <sub>t</sub>		3	mm	
Material		Low Carbon Steel	-	_	Air Gap





# Two-Wire, Differential, Vibration-Resistant Sensor with Speed and Direction Output

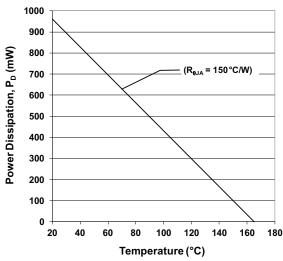
THERMAL CHARACTERISTICS: May require derating at maximum conditions; see Power Derating section

Characteristic	Symbol	Test Conditions*	Value	Unit
Package Thermal Resistance	$R_{\theta JA}$	Single layer PCB, with copper limited to solder pads	150	°C/W

<sup>\*</sup>Additional thermal information available on the Allegro website.

# Power Derating Curve V<sub>CC(max)</sub> V<sub>CC(max)</sub>

### Power Dissipation versus Ambient Temperature



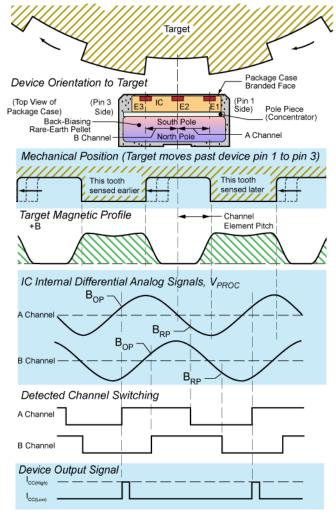


#### FUNCTIONAL DESCRIPTION

## **Sensing Technology**

The sensor IC contains a single-chip Hall-effect circuit that supports a trio of Hall elements. These are used in differential pairs to provide electrical signals containing information regarding edge position and direction of target rotation. The ATS699 is intended for use with ferromagnetic targets.

After proper power is applied to the sensor IC, it is capable of providing digital information that is representative of the magnetic features of a rotating target. The waveform diagrams in Figure 4 present the automatic translation of the target profiles, through their induced magnetic profiles, to the digital output signal of the sensor IC.



**Figure 4: Magnetic Profile** 

The magnetic profile reflects the features of the target, allowing the sensor IC to present an accurate digital output(-xSxxx variant shown).

#### **Direction Detection**

The sensor IC compares the relative phase of its two differential channels to determine which direction the target is moving. The relative switching order is used to determine the direction, which is communicated through the output protocol.

## **Data Protocol Description**

When a target passes in front of the device (opposite the branded face of the package case), the ATS699 generates an output pulse for each tooth of the target. Speed information is provided by the output pulse rate, while direction of target rotation is provided by the duration of the output pulses. The sensor IC can sense target movement in both the forward and reverse directions.

#### **FORWARD ROTATION**

As shown in panel A in Figure 5, when the target is rotating such that a tooth near the sensor IC—of -Fxxxx variant—passes from pin 1 to pin 3, this is referred to as forward rotation. This direction is opposite for the -Rxxxx variant. Forward rotation is indicated by output pulse widths of t<sub>w(FWD)</sub> (45 µs typical).

#### **REVERSE ROTATION**

As shown in panel B in Figure 5, when the target is rotating such that a tooth passes from pin 3 to pin 1, it is referred to as reverse rotation for the -Fxxxx variant. Reverse rotation is indicated by output pulse widths of  $t_{w(REV)}$  (90  $\mu$ s typical for -xxNxx variant, or 180  $\mu$ s typical for -xxWxx variant).

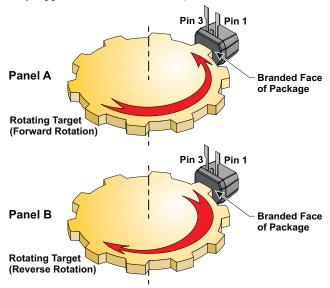


Figure 5: Target Rotation (F Variant Shown)



#### **TIMING**

As shown in Figure 6, the pulse appears at the output slightly before the sensed magnetic edge traverses the package branded face. For targets rotating from pin 3 to 1, this shift ( $\Delta$ fwd with R variants) results in the pulse corresponding to the valley with the sensed mechanical edge; for targets rotating from pin 1 to 3, the shift ( $\Delta$ rev) results in the pulse corresponding to the tooth with the sensed edge. Figure 7 shows pulse timing for F variants. The sensed mechanical edge that stimulates output pulses is kept the same for both forward and reverse rotation by using only one channel to control output switching.

## **Direction Validation**

For the -xxxxL variant, following a direction change in running mode, direction changes are immediately transmitted to the output.

For the -xxxxH variant, following a direction change in running mode, output pulses have a width of  $t_{w(ND)}$  until direction information is validated.

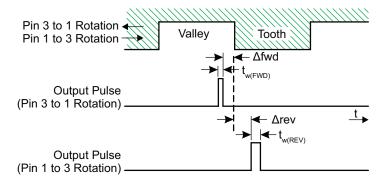


Figure 6: Output Protocol (-RSxxx Variant)

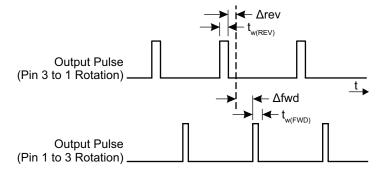


Figure 7: Output Protocol (-FDxxx Variant)

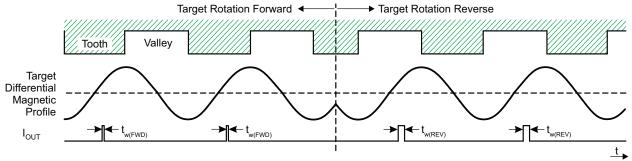


Figure 8: Running Mode Direction Change (-FSxxL Variant)

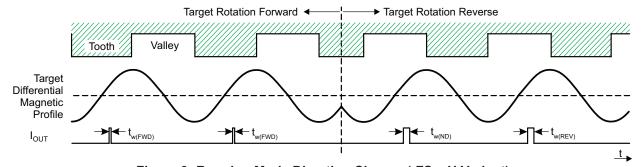


Figure 9: Running Mode Direction Change (-FSxxH Variant)



## **Start-Up Detection/Calibration**

When power is applied to the ATS699, the sensor IC internally detects the profile of the target. The gain and offset of the detected signals are adjusted during the calibration period, normalizing the internal signal amplitude for the air gap range of the device.

The Automatic Gain Control (AGC) feature ensures that operational characteristics are isolated from the effects of installation air gap variation.

Automatic Offset Adjustment (AOA) is circuitry that compensates for the effects of chip, magnet, and installation offsets. This circuitry works with the AGC during calibration to adjust  $V_{PROC}$ 

in the internal A-to-D range to allow for acquisition of signal peaks. AOA and AGC function separately on the two differential signal channels.

Direction information is available after calibration is complete.

For the –xxxBx variant, the output becomes active at the end of calibration. Figure 10 shows where the first output edges may occur for various starting target phases.

For the -xxxPx variant, output pulses of  $t_{w(ND)}$  are supplied during calibration.

Figure 11 shows where the first output edges may occur for various starting target phases.

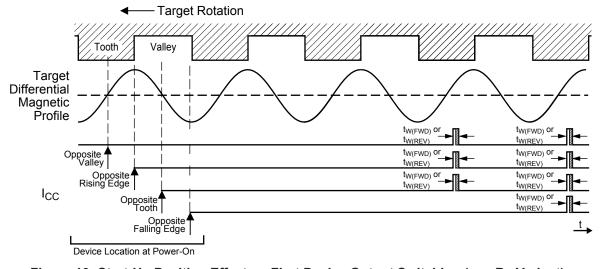


Figure 10: Start-Up Position Effect on First Device Output Switching (-xxxBx Variant)

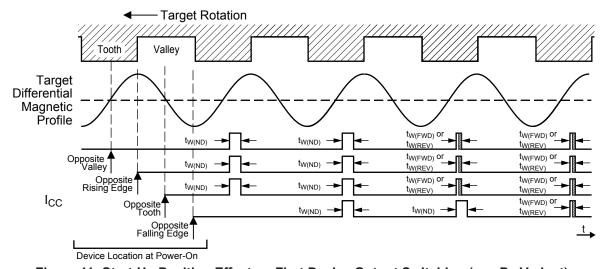


Figure 11: Start-Up Position Effect on First Device Output Switching (-xxxPx Variant)



#### Vibration Detection

Algorithms embedded in the IC's digital controller detect the presence of target vibration through analysis of the two magnetic input channels.

For the –xxxxL variant, the first direction change is immediately transmitted to the output. During any subsequent vibration, the output is blanked and no output pulses will occur for vibrations less than the specified vibration immunity. Output pulses contain-

ing the proper direction information will resume when direction information is validated on constant target rotation.

For the –xxxxH variant, in the presence of vibration, output pulses of  $t_{w(ND)}$  may occur or no pulses may occur, depending on the amplitude and phase of the vibration. Output pulses have a width of  $t_{w(ND)}$  until direction information is validated on constant target rotation.

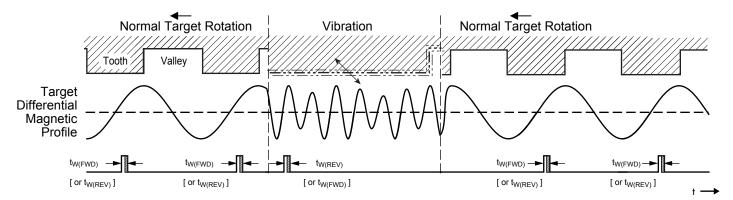


Figure 12: Output Functionality in the Presence of Running Mode Target Vibration (-xxxxL Variant)

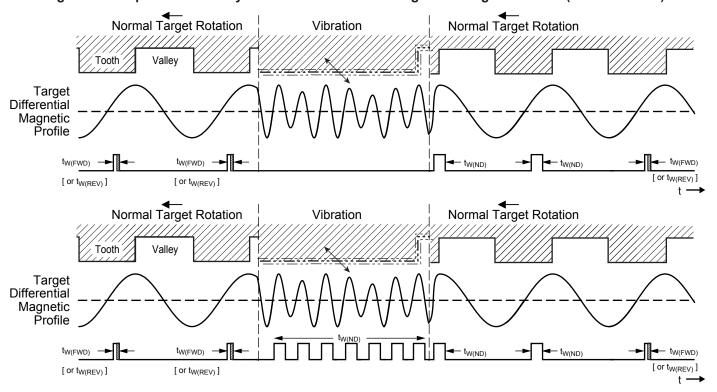


Figure 13: Output Functionality in the Presence of Running Mode Target Vibration (-xxxxH Variant)



#### POWER DERATING

The device must be operated below the maximum junction temperature of the device  $(T_{J(max)})$ . Under certain combinations of peak conditions, reliable operation may require derating supplied power or improving the heat dissipation properties of the application. This section presents a procedure for correlating factors affecting operating  $T_J$ . (Thermal data is also available on the Allegro website.)

The Package Thermal Resistance  $(R_{\theta JA})$  is a figure of merit summarizing the ability of the application and the device to dissipate heat from the junction (die), through all paths to the ambient air. Its primary component is the Effective Thermal Conductivity (K) of the printed circuit board, including adjacent devices and traces. Radiation from the die through the device case  $(R_{\theta JC})$  is relatively small component of  $R_{\theta JA}$ . Ambient air temperature  $(T_A)$  and air motion are significant external factors, damped by overmolding.

The effect of varying power levels (Power Dissipation,  $P_D$ ), can be estimated. The following formulas represent the fundamental relationships used to estimate  $T_I$ , at  $P_D$ .

$$P_D = V_{IN} \times I_{IN} \tag{1}$$

$$\Delta T = P_D \times R_{\theta,IA} \tag{2}$$

$$T_J = T_A + \Delta T \tag{3}$$

For example, given common conditions such as:  $T_A$ = 25°C,  $V_{CC}$  = 12 V,  $I_{CC}$  = 6 mA, and  $R_{\theta JA}$  = 150°C/W, then:

$$\begin{split} P_D &= V_{CC} \times I_{CC} = 12 \ V \times 6 \ mA = 72 \ mW \\ \Delta T &= P_D \times R_{\theta JA} = 72 \ mW \times 150 ^{\circ} C/W = 10.8 ^{\circ} C \\ T_J &= T_A + \Delta T = 25 ^{\circ} C + 10.8 ^{\circ} C = 35.8 ^{\circ} C \end{split}$$

A worst-case estimate,  $P_D(max)$ , represents the maximum allowable power level ( $V_{CC}(max)$ ,  $I_{CC}(max)$ ), without exceeding  $T_J(max)$ , at a selected  $R_{\theta JA}$  and  $T_A$ 

*Example*: Reliability for  $V_{CC}$  at  $T_A$ =150°C, package SN, using a single-layer PCB.

Observe the worst-case ratings for the device, specifically:  $R_{\theta JA} = 150^{\circ}\text{C/W}$ ,  $T_J(\text{max}) = 165^{\circ}\text{C}$ , and  $I_{CC(\text{mean})} = 13$  mA. (Note: For variant -xxWPx, at maximum target frequency,  $I_{CC(LOW)} = 8$  mA,  $I_{CC(HIGH)} = 16$  mA, and maximum pulse widths, the result is a duty cycle of 84% and thus a worst-case mean  $I_{CC}$  of 14.8 mA).

Calculate the maximum allowable power level, P<sub>D</sub>(max). First, invert equation 3:

$$\Delta T_{max} = T_J(max) - T_A = 165 \,^{\circ}C - 150 \,^{\circ}C = 15 \,^{\circ}C$$

This provides the allowable increase to  $T_J$  resulting from internal power dissipation. Then, invert equation 2:

$$P_D(max) = \Delta T_{max} \div R_{\theta,IA} = 15^{\circ}C \div 150^{\circ}C/W = 100 \text{ mW}$$

Finally, invert equation 1 with respect to voltage:

$$V_{CC}(est) = P_D(max) \div I_{CC}(max) = 100 \text{ mW} \div 14.8 \text{ mA} = 6.8 \text{ V}$$

The result indicates that, at  $T_A$ , the application and device can dissipate adequate amounts of heat at voltages  $\leq V_{CC}(est)$ .

Compare  $V_{CC}(est)$  to  $V_{CC}(max)$ . If  $V_{CC}(est) \leq V_{CC}(max)$ , then reliable operation between  $V_{CC}(est)$  and  $V_{CC}(max)$  requires enhanced  $R_{\theta JA}$ . If  $V_{CC}(est) \geq V_{CC}(max)$ , then operation between  $V_{CC}(est)$  and  $V_{CC}(max)$  is reliable under these conditions.



# Two-Wire, Differential, Vibration-Resistant Sensor with Speed and Direction Output

### PACKAGE OUTLINE DRAWING

## For Reference Only – Not for Tooling Use

(Reference Allegro DWG-0000429, Rev. 5)
Dimensions in millimeters – NOT TO SCALE
Dimensions exclusive of mold flash, gate burs, and dambar protrusions
Exact case and lead configuration at supplier discretion within limits shown

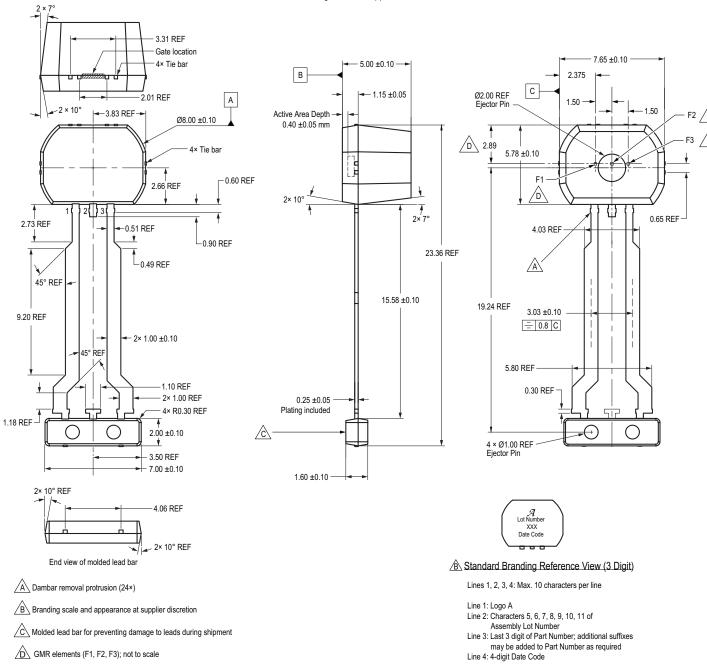


Figure 15: Package SN, 3-Pin SIP



## Two-Wire, Differential, Vibration-Resistant Sensor with Speed and Direction Output

#### **Revision History**

Number	Date	Description
_	March 2, 2017	Initial release
1	March 9, 2017	Updated Thermal Characteristics and Power Derating sections
2	May 19, 2017	Updated Selection Guide (page 2, clarified Forward Rotation Direction option).
3	February 27, 2019	Minor editorial updates
4	March 18, 2020	Minor editorial updates
5	March 9, 2022	Updated package drawing (page 14)

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